Si5040

10 Gbps XFP Transceiver with Jitter Attenuator

Features

Complete, high-performance, low-power, 10 Gbps XFP transceiver featuring independent CDRs, DSPLL®-based jitter-attenuating CMUs, and data retimers in both transmit and receive directions.

- DSPLL-based, jitter-attenuating CMUs in both transmit and receive directions
- Frequency-agile jitter filtering from 9.8 to 11.35 Gbps (continuous)
- Compliant to XFP specifications and jitter specifications for telecom (SONET/SDH, OTU-2) and datacom (10 GbE/10 GbE+FEC and 10 GFC/10 GFC+FEC) applications
- Supports referenceless operation
- Integrated limiting amplifier provides better than 8 mV receiver sensitivity
- User-programmable receiver loss-of-signal (LOS) detector

- Transmitter jitter generation 2.5 mUI$_{rms}$ (typical)
- Automatic slicing level adjustment with optional programmable override
- Programmable sample phase adjustment
- Line loopback, XFI loopback, pattern generation, and pattern check test capabilities
- 1.8/3.3 V or single 1.8 V supply
- 575 mW (typ) power dissipation
- 5x5 mm LGA package
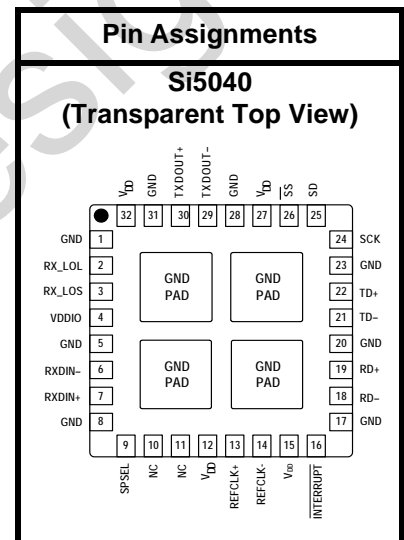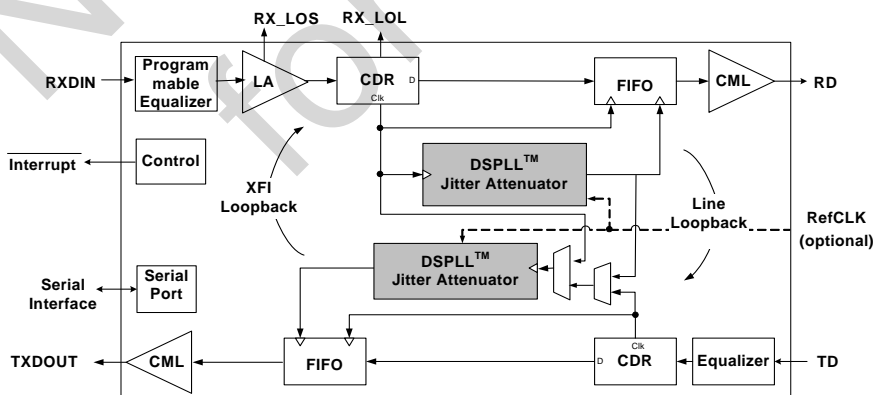- Serial microcontroller interface control

Applications

- XFP telecom modules
- XFP datacom modules
- Optical test equipment

- Jitter-attenuation and signal regeneration of 10 Gbps serial signal on line cards

Ordering Information:
See page 103.

Pin Assignments

**Si5040
(Transparent Top View)**

Description

The Si5040 is a complete, low-power, high-performance XFP transceiver suitable for multiple XFP module types, from short-reach datacom to long-reach telecom applications. The Si5040 integrates a rate-agile, programmable-bandwidth, jitter-attenuating CMU in the transmit direction, which significantly attenuates jitter present at the XFI interface and on the applied reference clock, removing the need for an external jitter cleanup circuit. The device supports referenceless operation or operation with a synchronous or asynchronous reference clock. The device can be completely configured through a serial microcontroller interface. The Si5040 is compliant with all XFP requirements in both datacom and telecom applications. The Si5040 is packaged in a 5x5 mm LGA package and dissipates 575 mW (typ).
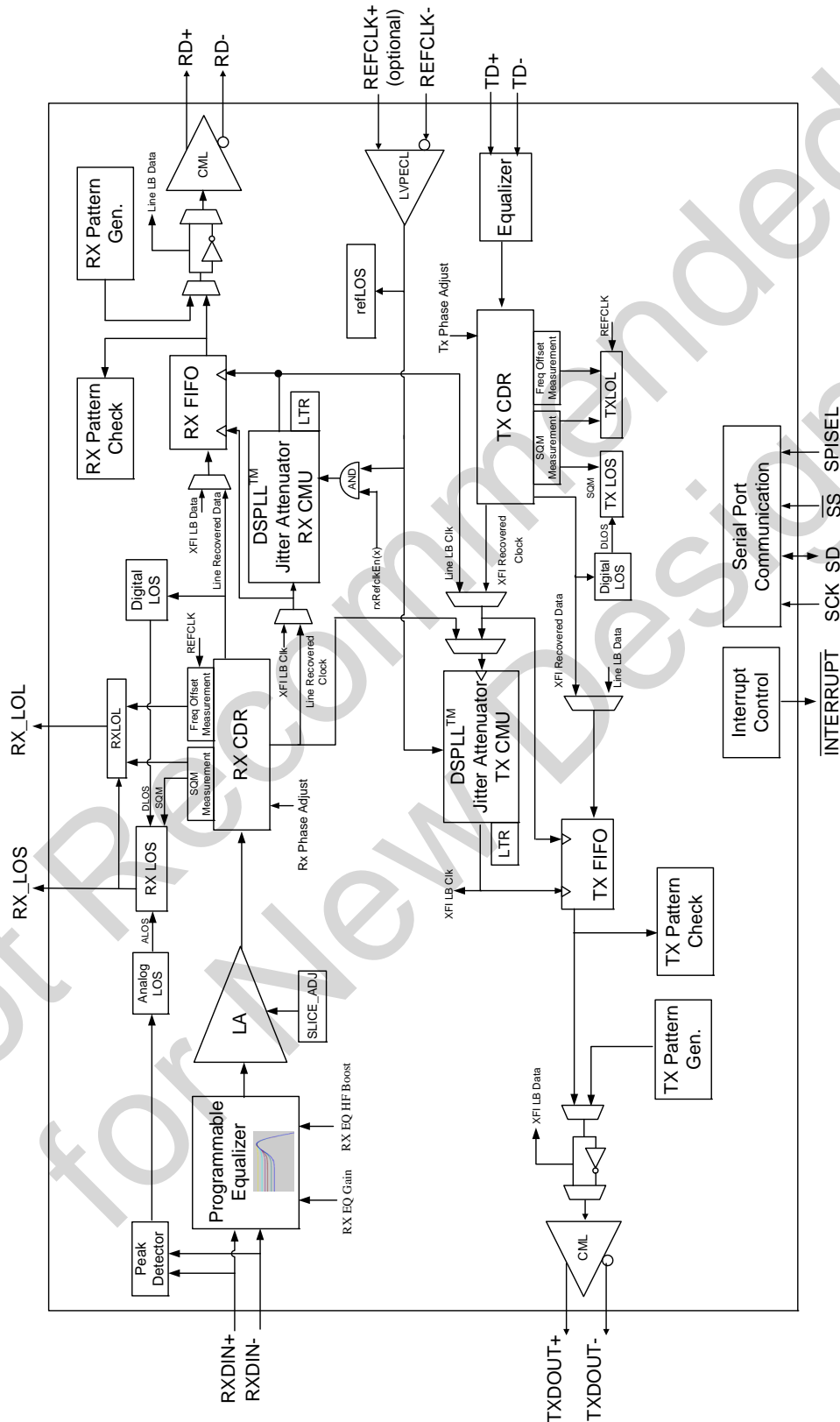
Functional Block Diagram

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1. Si5040 Detailed Block Diagram

2. Electrical Specifications

Table 1. Recommended Operating Conditions

Parameter	Symbol	Test Condition	Min[1]	Typ	Max[1]	Unit
Ambient Temperature	T_A		−40	25	95	°C
Supply Voltage	V_{DD}[2]		1.62	1.80	1.89	V
LVTTL I/O Supply Voltage	V_{DDIO}		1.62	—	3.63	V

Notes:
1. All minimum and maximum specifications are guaranteed and apply across the recommended operating conditions. Typical values apply at nominal supply voltages and an operating temperature of 25 °C unless otherwise stated.
2. Maximum power ramp-up time to reach $V_{DD} \geq 1.62$ V should be less than 50 ms. Any ramp-up time slower than 50 ms will require a software reset to ensure proper calibration of all internal circuits.
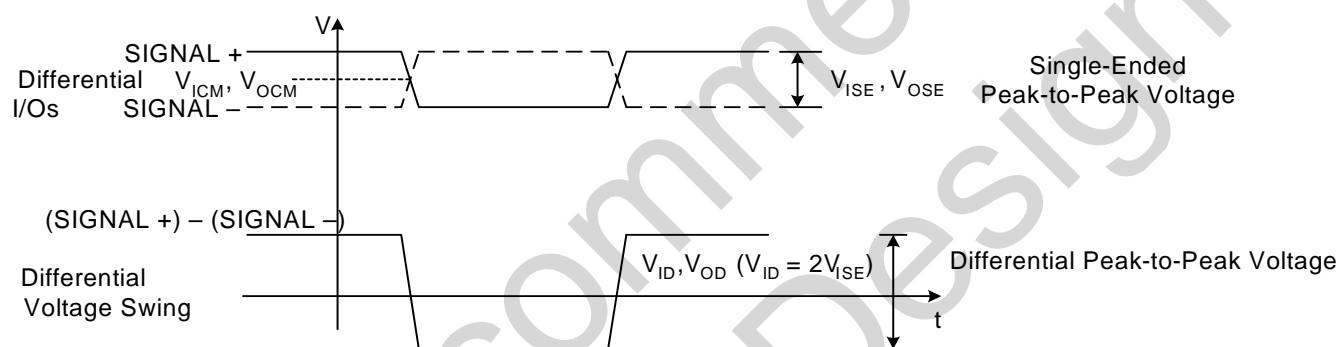
Figure 1. Voltage Measurement
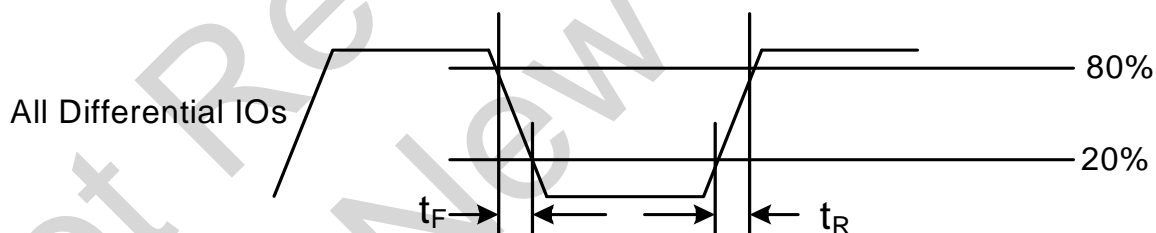
Figure 2. Rise/Fall Time Measurement

Table 2. DC Characteristics

(V_{DD} = 1.8 V +5%/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Supply Current	I_{DD}		—	—	370	mA
Power Dissipation	P_D	V_{DD} = 1.89 V[1] V_{DD} = 1.8 V[1]	— —	— 575	700 —	mW
Differential Input Voltage Swing (RXDIN) (at BER 10^{-12})	V_{ID}	Figure 1 with Receive Equalizer bypassed	8	—	1000	mV$_{PPD}$
Common Mode Output Voltage (TXDOUT)	V_{OCM}	Figure 1	—	1.3 – V_{OD}/2	—	V
Differential Output Voltage Swing (TXDOUT), Differential Peak-to-Peak OutLevel[2:0] = 111 110 101 100 011 010 001 000	V_{OD}	Figure 1 R_L = 100 Ω	800 700 600 500 400 300 200 100	— — — — — — — —	— — — — — — — —	mV$_{PPD}$
LVPECL Input Voltage Swing, Differential Peak-to-Peak (REFCLK)[2]	V_{ID}	Figure 1	250	—	2400	mV$_{PPD}$
Differential Input Impedance (TD, REFCLK, RXDIN)	R_{IN}		—	100	—	Ω
LVTTL Input Voltage Low	V_{IL}	V_{DDIO} = 3.3 V[3]	–0.3	—	0.8	V
LVTTL Input Voltage High	V_{IH}	V_{DDIO} = 3.3 V[3]	2.0	—	V_{DDIO} + 0.3	V
LVTTL Input Impedance	R_{IN}		10	—	—	kΩ
LVTTL Output Voltage Low (I_{OUT} = 2 mA)	V_{OL}	V_{DDIO} = 3.3 V[3]	—	—	0.4	V
LVTTL Output Voltage High (I_{OUT} = 2 mA)	V_{OH}	V_{DDIO} = 3.3 V[3]	2.4	—	—	V
Differential Output Impedance (RD, TXDOUT)	R_{OUT}		—	100	—	Ω
Termination Mismatch (RD)			—	—	5	%
Termination Mismatch (TD)			—	—	5	%

Notes:
1. TX CMU Mode 0.
2. REFCLK must be ac-coupled. For CMU Mode 1 or TX LTR operation, the minimum input swing should be 650 mV$_{PPD}$.
3. V_{DDIO} = 1.8 V, not characterized.

Table 2. DC Characteristics (Continued)

(V_{DD} = 1.8 V +5%/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
I²C Bus Lines (SD, SCK)						
Input Voltage Low	V_{ILI2C}		—	—	0.3 x V_{DDIO}	V
Input Voltage High	V_{IHI2C}		0.7 x V_{DDIO}	—	—	V
Input Current	I_{II2C}	V_{IN} = 0.1 x V_{DDIO} to 0.9 x V_{DDIO}	–10	—	10	µA
Hysteresis of Schmitt trigger inputs	V_{HYSI2C}	V_{DDIO} = 1.8 V	0.1 x V_{DD}	—	—	V
		V_{DDIO} = 3.3 V	0.05 x V_{DD}	—	—	V
Output Voltage Low	V_{OLI2C}	V_{DDIO} = 1.8 V I_O = 3 mA	—	—	0.2 x V_{DD}	V
		V_{DDIO} = 3.3 V I_O = 3 mA	—	—	0.4	V

Notes:
1. TX CMU Mode 0.
2. REFCLK must be ac-coupled. For CMU Mode 1 or TX LTR operation, the minimum input swing should be 650 mV_{PPD}.
3. V_{DDIO} = 1.8 V, not characterized.

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Table 3. AC Characteristics–RXDIN (Receiver Input)

(V_{DD} = 1.8 V +5/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
RX Path Data Rate			9.80	9.95	11.35	Gbps
Input Return Loss (RXDIN)		< 2 GHz	15	—	—	dB
		5 GHz	10	—	—	dB
		10 GHz	5	—	—	dB
ALOS Range		Analog Mode	10	—	400	mV$_{PPD}$
ALOS Step Size		Analog Mode	—	1	—	mV$_{PPD}$
ALOS Relative Accuracy		Analog Mode	1	—	—	mV
DLOS Range		Consecutive Digits Mode	0.5	—	100	µs
DLOS Accuracy		Consecutive Digits Mode	0.5	—	—	µs
ALOS Hysteresis (Programmable 0.4 dB steps)		Analog Mode	0	—	6	dB
Slice Voltage Range		mode = absolute or proportional	–240	—	240	mV$_D$
Slice Voltage Error		Max error from the programmed absolute slice voltage			±20	%
Sample Phase Range			–12	—	12	ps
Jitter Tolerance OC-192 BER = 10^{-12}	J$_{TOL(PP)}$	f = 2.0 kHz	15	30	—	UI$_{PP}$
		f = 20 kHz	1.5	3.0	—	UI$_{PP}$
		f = 400 kHz	1.5	3.0	—	UI$_{PP}$
		f = 4 MHz	0.4	—	—	UI$_{PP}$
		f = 80 MHz	0.4	—	—	UI$_{PP}$
Acquisition Time (Default mode) REFCLK Referenceless	T$_{AQ}$	Register 86, bit 6:3 = 8 decimal Register 68, bit 3:0 = 2 decimal Register 67, bit 6:2 = 17 decimal	— —	— —	50 50	ms ms
Frequency Difference at which Receive PLL Goes Out of Lock	LOL		800	—	—	ppm
Frequency Difference at which Receive PLL Goes into Lock	LOCK		—	—	200	ppm

Table 4. AC Characteristics—RD (Receiver Output)

(V_{DD} = 1.8 V +5/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Common Mode AC Output Voltage			—	—	15	mV_{RMS}
Output Rise and Fall Times (RD)	t_R, t_F	Figure 2	24	—	—	ps
Differential Output Return Loss	SDD22	0.05–0.1 GHz	20	—	—	dB
		0.1–5.5 GHz	8	—	—	dB
		5.5–12 GHz	*	—	—	dB
Common Mode Output Return Loss	SCC22	0.1–15 GHz	3	—	—	dB
Deterministic Jitter	DJ	>4 MHz. See Appendix E1 in the XFP specification.	—	—	0.09	UI_{PP}
Total Jitter	TJ		—	—	0.17	UI_{PP}
Eye Mask	X1	See Figure 3.	—	0.08	—	UI
Eye Mask	X2	See Figure 3.	—	0.33	—	UI
Eye Mask	Y1	See Figure 3.	190	—	—	mV
Eye Mask	Y2	See Figure 3.	—	—	385	mV
Jitter Transfer Bandwidth (Programmable) cmuBand-width[3:0] (Register 6, bits 7:4) 0100	J_{BW}	9.95 Gbps	—	380	760	kHz

*Note: Differential return loss given by equation SDD22 (dB) = 8 – 20.66 Log_{10}(f/5.5), with f_{in} GHz.

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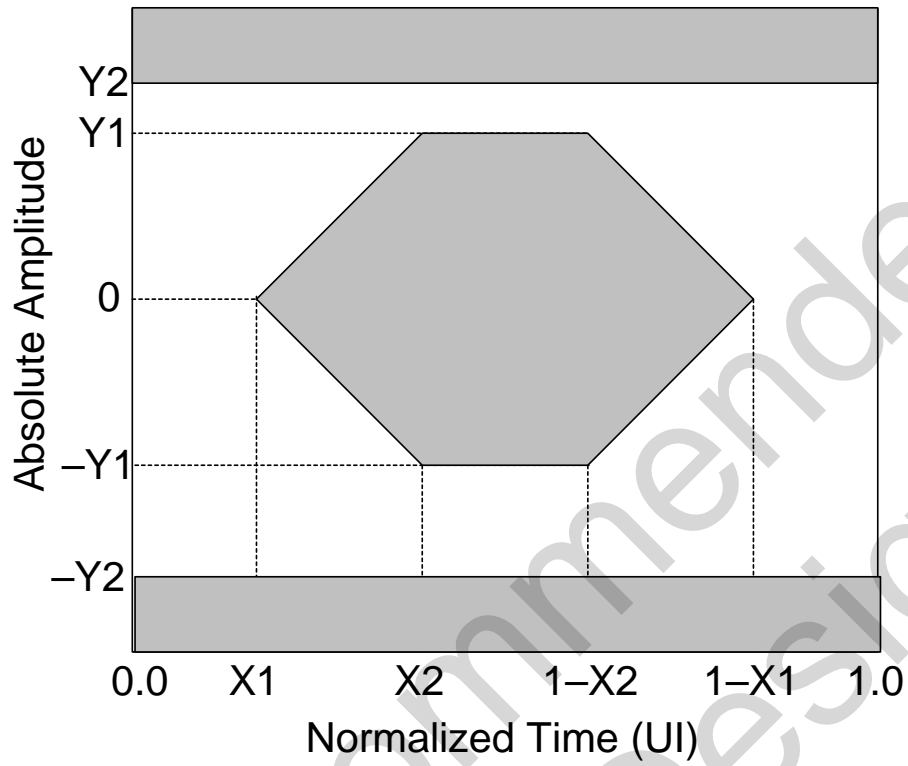
Figure 3. Receiver Differential Output Mask (RD)

Table 5. AC Characteristics—TXDOUT (Transmitter Output)

(V_{DD} = 1.8 V +5/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Output Rise + Fall Times	t_R, t_F	Figure 2	20	25	30	ps
Output Return Loss		400 kHz–10 GHz 10 GHz–16 GHz	6 4	— —	— —	dB dB
Random RMS Jitter Generation, TXDOUT[1]	$J_{GEN(rms)}$	OC-192, CMU mode 0	—	2.8	4.6	mUI$_{RMS}$
Total Peak-to-Peak Jitter Generation, TXDOUT[1]	$J_{GEN(PP)}$	OC-192, CMU mode 0	—	36	60	mUI$_{PP}$
Jitter Transfer Bandwidth (Programmable) CMU bandwidth [3:0] (Register 134, bits 7:4) 0000 0001 0010 0100 0101 0110	J_{BW}	9.95 Gbps	— — — — — — —	 180 1.37 Not supported 380 Not supported Not supported	 220 1.76 760 	 Hz kHz kHz kHz kHz kHz
Jitter Transfer Peaking		< 120 kHz	—	—	0.03	dB
Acquisition Time REFCLK Referenceless	T_{AQ}		 — —	 — —	 50 61	 ms ms
Input Reference Clock Frequency[2]	RC_{FREQ}	ref clk /16 mode ref clk /64 mode	618.75 154.687	622 155	709.4 177.35	MHz
Input Reference Clock Duty Cycle	RC_{DUTY}		40	—	60	%
Input Reference Clock Frequency Tolerance	RC_{TOL}		–100	—	100	ppm

Notes:
1. PRBS31 or SONET framed PRBS31 data. The integrated CMU filters out SONET framing effects.
2. Input reference clock frequency can be either Baud rate/16 or Baud rate/64 ±100 ppm. The typical and maximum numbers specified here correspond to /16 or /64 of the typical and maximum data rate that the device supports.

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Table 6. AC Characteristics–TD (Transmitter Input)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
TX Path Data Rate			9.80	9.95	11.35	Gbps
Differential Input Return Loss	SDD11	0.05–0.1 GHz	20	—	—	dB
		0.1–5.5 GHz	8	—	—	dB
		5.5–12 GHz	See Note [1]	—	—	dB
Common Mode Input Return Loss (TD)	SCC11	0.1–15 GHz	3	—	—	dB
Total Non-DDJ[2]				—	0.45	UI_{PP}
Total Jitter		TJ		—	0.65	UI_{PP}
Sinusoidal Jitter Tolerance[3]		SJ	—	See Figure 4	—	UI_{PP}
Eye Mask (See Figure 5)	X1		—	—	0.325	UI
Eye Mask (See Figure 5)	Y1		50	—	—	mV
Eye Mask (See Figure 5)	Y2		—	—	525	mV

Notes:
1. Return loss is given by the following equation: SDD11(dB) = 8–20.66Log10(f/5.5), with f in GHz.
2. Total jitter less ISI.
3. The jitter tolerance given in Figure 5 is in addition to the random and deterministic jitter given in this table.

Table 7. CMU Timing Modes

Mode	Description	Typical Jitter Gen.	Recommended CMU Jitter Transfer Bandwidth Setting
0	No Reference Clock or Asynchronous Reference Clock[1,2]	3.5 mUI_{rms}	380 kHz
1	Clean, synchronous[3]	2.5 mUI_{rms}	380 kHz
2	Clean, asynchronous[1,3]	2.5 mUI_{rms}	1.37 kHz

Notes:
1. Reference clock with frequency equal to Baud rate /64 ±100 ppm and phase noise as defined in XFP Specification 3.1, Section 3.9.
2. Since the default bandwidth for this mode is 100 kHz, Register 134 [7:4] should be written to a "4" to set the bandwidth to 380 kHz.
3. Clean reference clock with frequency equal to exactly Baud rate /64 and phase noise as defined in XFP Specification 3.1, Section 3.9.1. REFCLK input amplitude >650 mVppd.

Si5040 TD Jitter Tolerance (Typ)
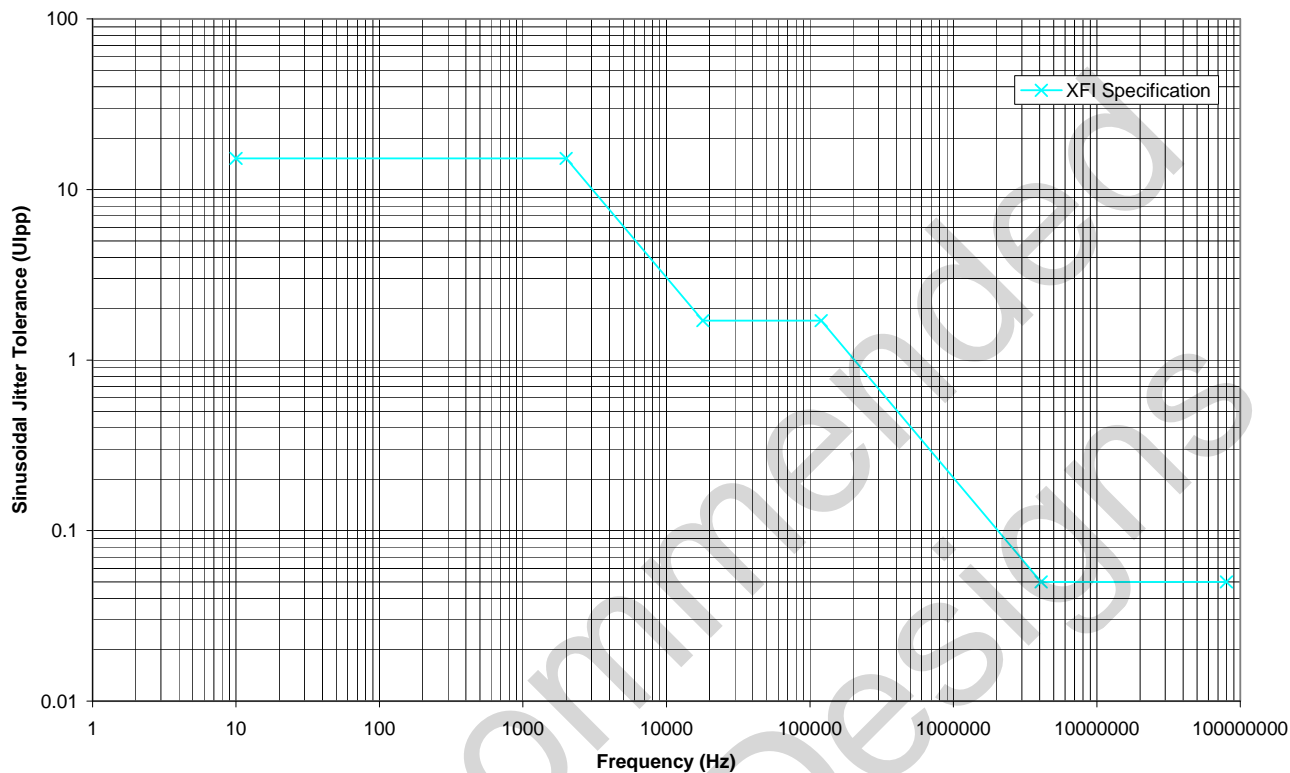
Figure 4. XFI Sinusoidal Jitter Tolerance (UI$_{PP}$)
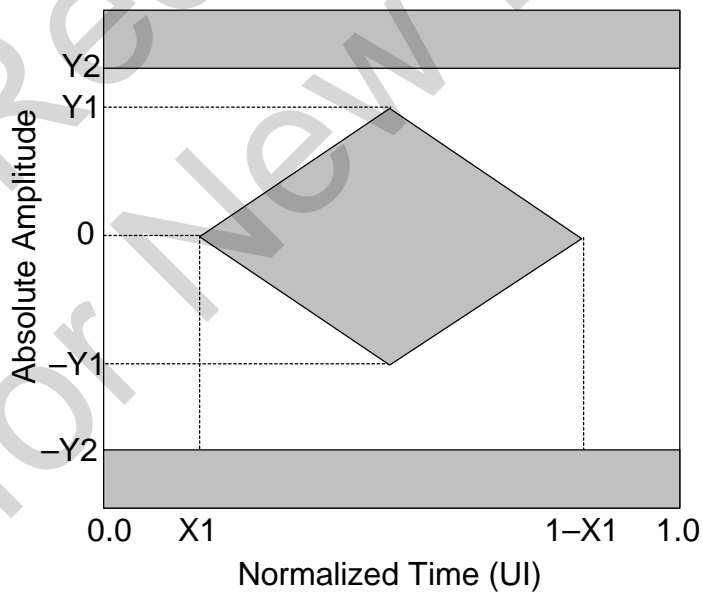
Figure 5. Transmitter Differential Input Mask (TD)

Table 8. AC Characteristics—I^{2}C Bus Lines (SD, SCK)

(V_{DD} = 1.8 V +5/–10%, T_A = –40 to 95 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Pin Capacitance	C_{II2C}		—	—	10	pF

Table 9. Switching Characteristics—Serial Microcontroller Interface[2]

V_{DD} = 1.8 V +5/–10%, V_{DDIO} = 3.3 V ±10%, T_A = –40 to 95 °C, C_L = 20 pF

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Cycle Time SCK	t_c		100	—	—	ns
Rise Time, SCK	t_r	20–80%	—	—	25	ns
Fall Time, SCK	t_f	20–80%	—	—	25	ns
Low Time, SCK	t_{lsc}	20–20%	30	—	—	ns
High Time, SCK	t_{hsc}	80–80%	30	—	—	ns
Delay Time, SCK Fall to SD Active	t_{d1}		—	—	25	ns
Delay Time, SCK Fall to SD Transition	t_{d2}		—	—	25	ns
Delay Time, $\overline{SS}$ Rise to SD Tri-state[1]	t_{d3}		—	—	25	ns
Setup Time, $\overline{SS}$ to SCK Fall	t_{su1}		25	—	—	ns
Hold Time, $\overline{SS}$ to SCK Rise	t_{h1}		20	—	—	ns
Setup Time, SD to SCK Rise	t_{su2}		25	—	—	ns
Hold Time, SD to SCK Rise	t_{h2}		20	—	—	ns
Delay Time between Slave Selects	t_{cs}		25	—	—	ns

Notes:
1. SD is designed to be tristated by the release of the chip select signal (the rising edge of the $\overline{SS}$).
2. All timing is referenced to the 50% level of the waveform unless otherwise noted. Input test levels are V_{IH} = V_{DD} –0.4 V, V_{IL} = 0.4 V
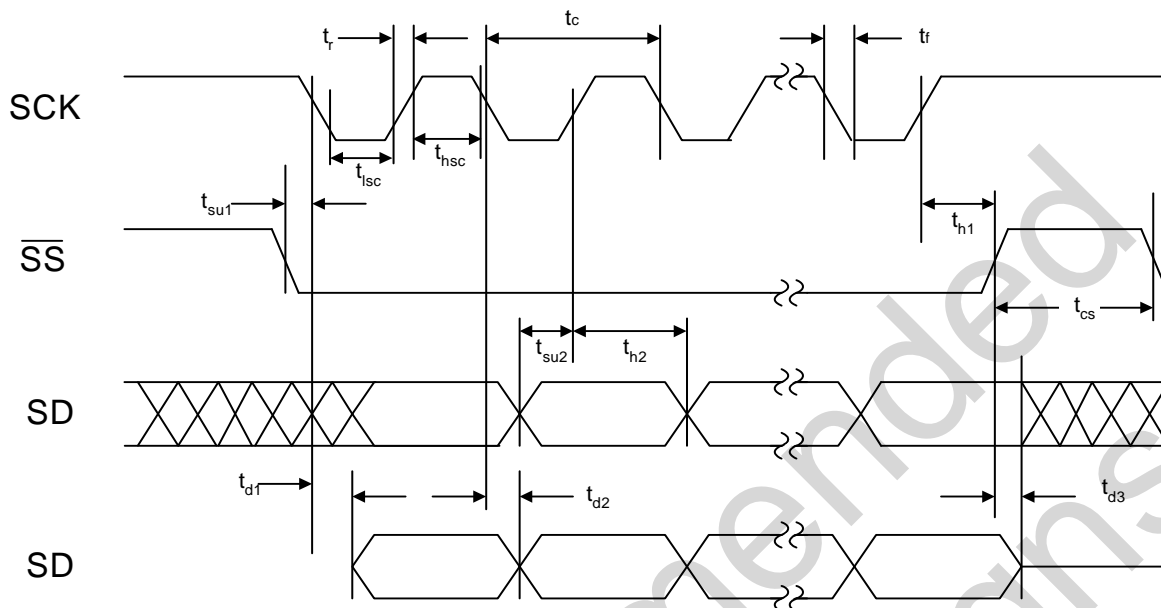
Figure 6. Serial Microcontroller Interface Timing Diagram
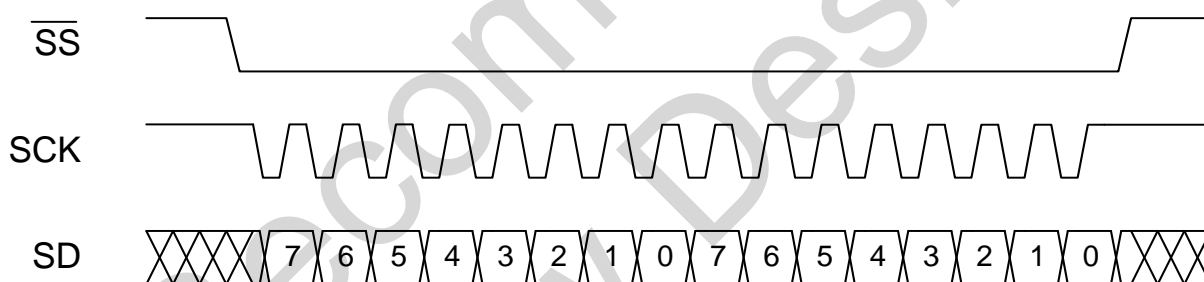
Figure 7. SPI-Like Interface Write/Set Address Command

Read Command   Read Data
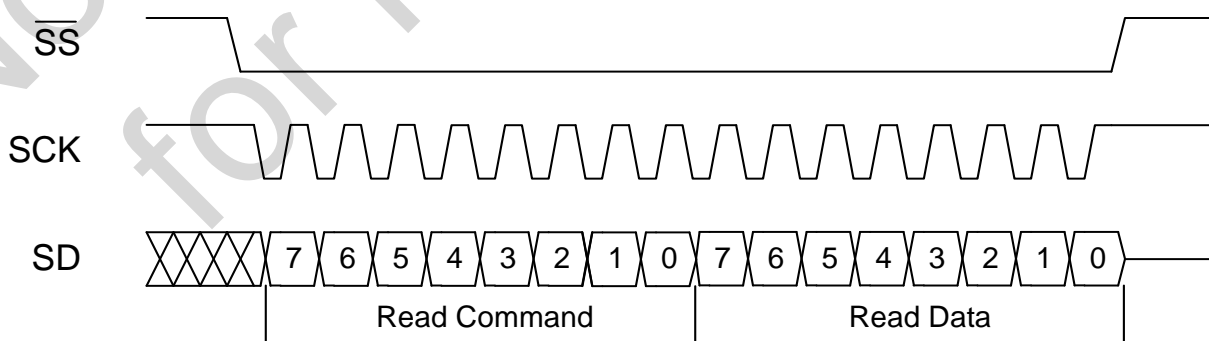
Figure 8. SPI-Like Interface Read Command

Table 10. Thermal Characteristics

Parameter	Symbol	Test Condition	Value	Unit
Thermal Resistance Junction to Ambient	φ_{JA}	Still Air	50	°C/W

Table 11. Absolute Maximum Ratings*

Parameter	Symbol	Value	Unit
DC Supply Voltage	V_{DD}	–0.5 to 1.98	V
LVTTL Supply Voltage	V_{DDIO}	–0.5 to 3.8	V
Differential Input Voltages	V_{DIF}	–0.3 to (V_{DD}+ 0.3)	V
Maximum Current any Output PIN		±50	mA
Operating Junction Temperature	T_{JCT}	–55 to 150	°C
Storage Temperature Range	T_{STG}	–55 to 150	°C
ESD HBM Tolerance (100 pf, 1.5 kΩ)	High-speed pins	2	kV
	All except high-speed pins	2	kV
*Note: Permanent device damage can occur if the absolute maximum ratings are exceeded. Restrict functional operation to the conditions specified in the operational sections of this data sheet. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.			

3. Typical Application Schematic
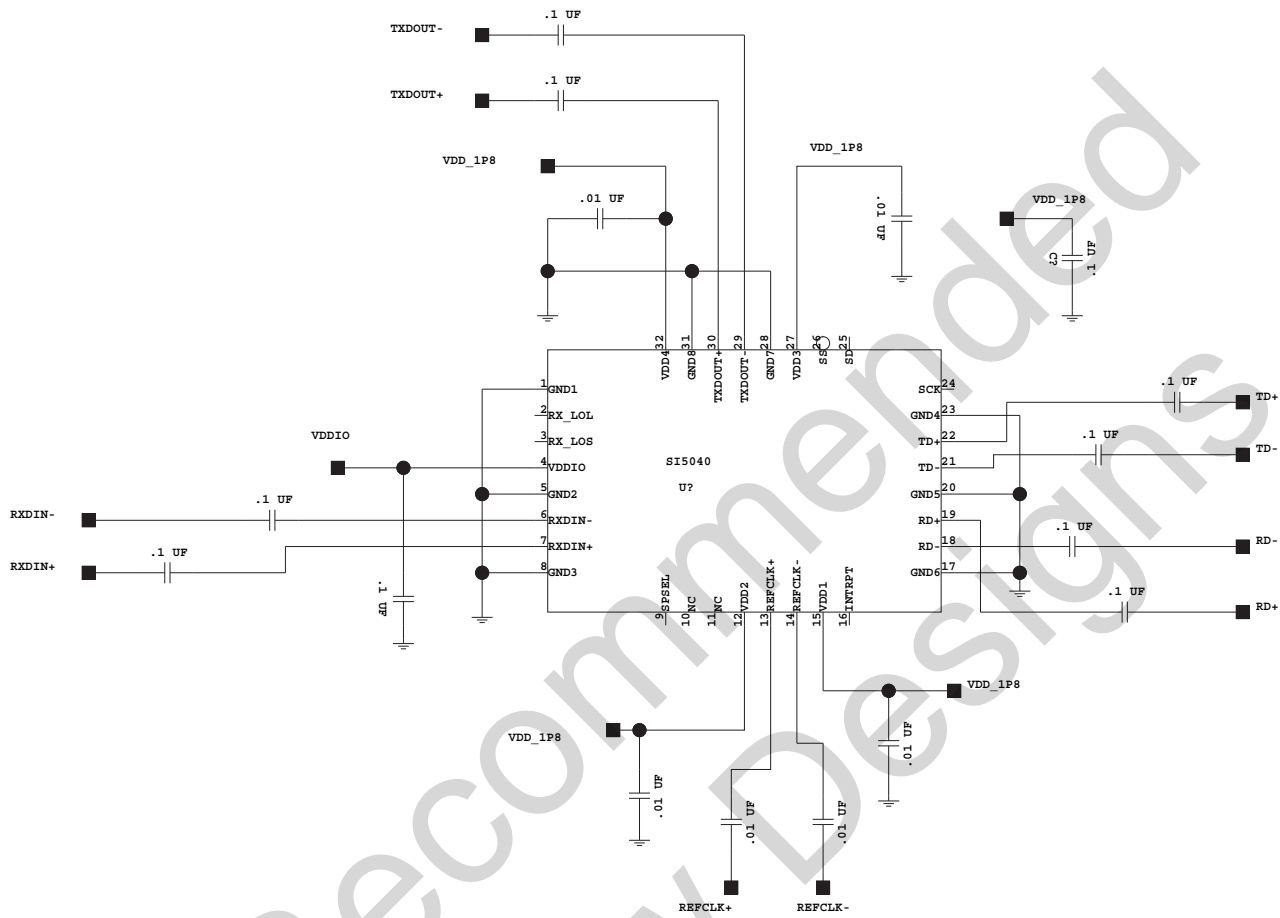
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4. Functional Description

The Si5040 XFP transceiver is a single-chip, bidirectional signal conditioner for use in XFP modules as defined by the XFP multi-source agreement. The Si5040 includes independent clock and data recovery units (CDRs) and frequency-agile, jitter-attenuating clock multiplier units (CMUs) in both receive and transmit directions. The receive path includes a limiting amplifier and a programmable equalizer for direct connection to an optical receiver trans-impedance amplifier. The transmit path includes an equalizer for direct connection to the XFI channel.

The device provides data-agnostic operation over a continuous range of data rates from 9.8 to 11.35 Gbps.

Typical data rates and associated applications include the following:

- 9.8304 Gbps: Common Public Radio Interface (CPRI)
- 9.95 Gbps: SONET OC-192, SDH STM-64, 10 Gbps Ethernet WAN PHY
- 10.31 Gbps: 10 Gbps Ethernet LAN PHY
- 10.52 Gbps: 10 Gbps FibreChannel (10 GFC)
- 10.70 Gbps: OTN OTU-2 (G.709)
- 11.09 Gbps: 10 Gbps Ethernet LAN PHY with 255/237 FEC coding
- 11.32 Gbps: 10 Gbps FibreChannel with 255/237 FEC coding

Serial control and status monitoring is supported with either an SPI-like or I^2C serial interface.

5. Receiver

The Si5040 receiver includes a programmable equalizer, a high-sensitivity limiting amplifier, clock and data recovery unit (CDR), and a FIFO/retimer function.

5.1. Receive Equalizer

The RX equalizer is a programmable equalizer designed to boost the high-frequency components while attenuating the low-frequency components. Figure 9 illustrates a typical frequency response of the RX equalizer when its capacitor setting (or high-frequency boost, RxEqHFBoost at Register 85, Bit [7:5]) is set to the maximum value and its resistor setting (or low-frequency gain/attenuation, RxEqGain at Register 84, Bit [7:5]) is changed to achieve different low-frequency attenuations.
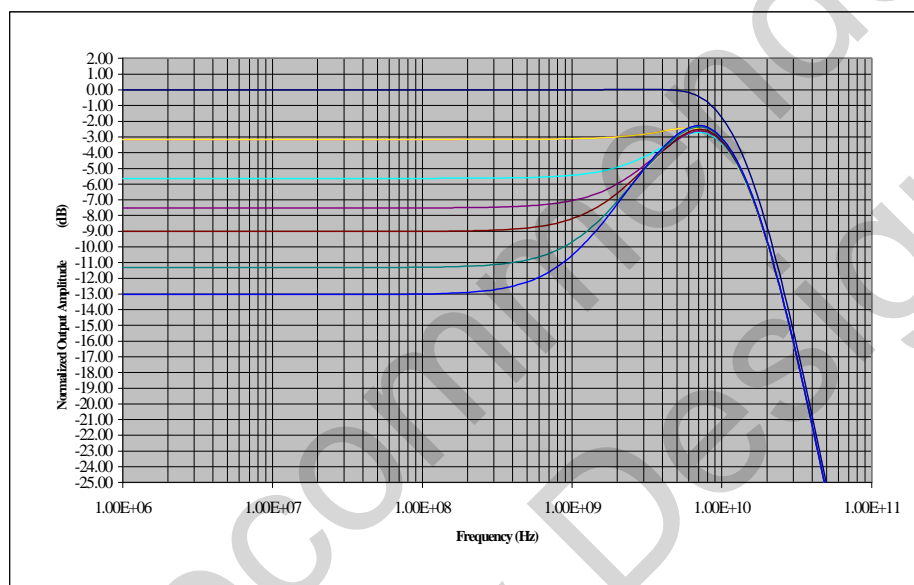
**Figure 9. Typical RX Equalizer
Frequency Response**

To optimize the equalizer settings for a given application, the following procedure is recommended:

- Set RxEqHFBoost at Register 85, Bit [7:5] to 7 (Default = 4), and adjust RxEqGain at Register 84, Bit [7:5] (Default = 5), to achieve the best performance.
- Further adjust RxEqHFBoost to increase the performance further.

Note that setting RxEqGain to 0 dB bypasses the equalizer.

5.2. Limiting Amplifier

The Si5040 incorporates a high-sensitivity differential limiting amplifier with sufficient gain to allow direct connection to a transimpedance amplifier. The amplifier has a guaranteed sensitivity of 8 mVppd.

5.3. Receive Amplitude Monitor

The Si5040 limiting amplifier includes circuitry that monitors the amplitude of the receiver differential input signal (RXDIN). The amplitude value can be read from the Peakdet register (Register 16). The receiver signal amplitude monitoring circuit is also used in the generation of the loss-of-signal alarm (LOS).

5.4. Receiver Loss of Signal Alarm (LOS)

The Si5040 receiver generates a loss-of-signal alarm when the input signal fails to meet the selected programmable condition for loss of signal. The programmable LOS mode is controlled in the RxlosCtrl register (Register 10). The programmable modes are Analog Loss Of Signal (ALOS), Digital Loss of Signal (DLOS), and Signal Quality Monitor (SQM). As shown in Figure 15, one or more of these alarm signals is logically combined into the Receiver Loss of Signal (RX_LOS) alarm pin (Pin 3). The state of the RX_LOS alarm pin will also be reflected in the LOS bit in the RxAlarmStatus register (Register 9). LOS may also be configured to generate an interrupt. The status of the LOS interrupt bit may be read in the RxintStatus register (Register 5). The status of the various LOS modes is stored in the losStatus register (Register 11).

An ALOS alarm occurs when the peak-to-peak signal amplitude on the RXDIN input is below the threshold value set in the aLosThresh register (Register 12). The level may be set from 10 to 400 mV in 1 mV increments. The amount of hysteresis applied to ALOS is set in the aLosThresh2 register (Register 13) up to a maximum of "aLoSThresh+" 6 dB. The default value is "aLoSThresh+" 3 dB. The analog LOS function is enabled by setting bit 0 in the RxLosCtrl register (Register 10). Note that the peak-to-peak detector and ALOS detection are designed to work with pseudo-random dc-balanced data band limited to <2 GHz.

A DLOS alarm occurs when the receive bit stream consists of a run length of 1s or 0s greater than the value loaded in the RxdLosAssertThresh register (Register 17). DLOS will remain asserted until the data shows activity for a time greater than that loaded in the RxdLOSClearThresh register (Register 18). Refer to Figure 10 and Figure 11 for algorithm flowcharts.

An SQM alarm occurs when the estimate of the receive signal quality falls below the value loaded in RxsqmThresh (Register 26). The Signal Quality Monitor measures the magnitude of the horizontal eye opening of the received signal. The SQM value can be read from the RxsqmValue register (Register 25). An SQM alarm will assert if the RxsqmEn bit has been set in the RxsqmConfig register (Register 26). SQM hysteresis is set in the RxsqmDeassertThresh register (Register 27).

```
        ┌─────────────────────┐
        │  CID (Continuous    │◄──────────────┐
        │  Identical Digits)  │               │
        │     Monitor         │               │
        └─────────────────────┘               │
                   │                           │
                   ▼                           │
            ◇─────────────────◇                │
      Is CID ≥ (dLosAssertThresh *5 +2) *1024    No
            ◇─────────────────◇────────────────┘
                   │
                  Yes
                   │
                   ▼
                Set dLos
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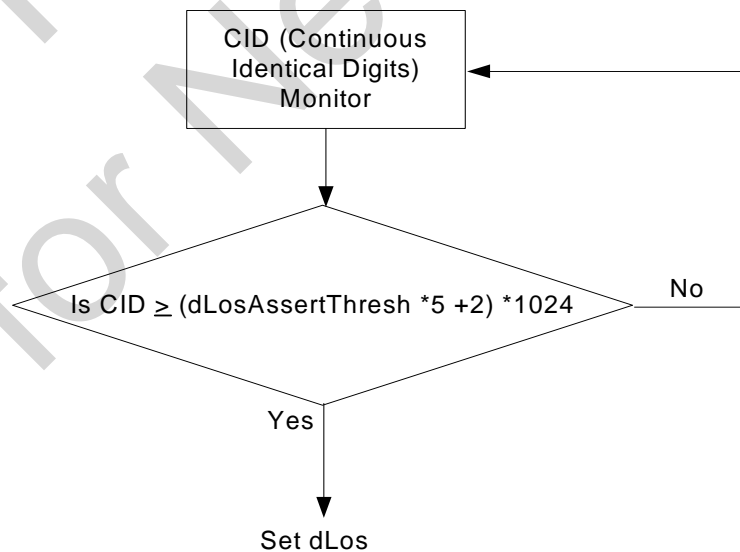
Figure 10. Algorithm to Set dLOS
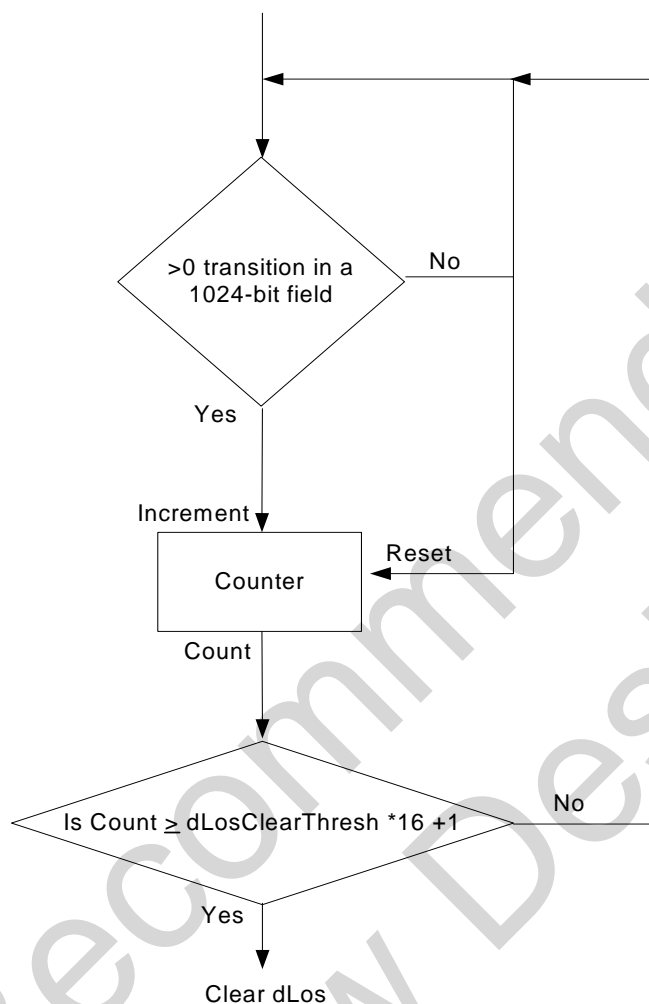
Figure 11. Algorithm to Clear dLOS

The receiver may be programmed to perform any of the following consequent actions upon declaring RX_LOS:

1. Lock the receiver to the applied reference clock (lock to reference): ltrOnLOS bit in Register 7.

2. Assert receiver loss of lock (LOL): lolOnLOS bit in Register 7.

3. Disable (squelch) the receive data output (RD): SquelchOnRxLOS bit in Register 28.

4. Generate a clock pattern at the receive data output (RD): clkOnLOS bit in Register 28.

For different combinations of ltrOnLOS and lolOnLOS settings in Register 7, the device may behave differently in the CDR lock acquisition process. Refer to Figure 12 and Figure 13 for more details on CDR and VCO behaviors upon declaring LOS. ltrOnLOS = 0, lolOnLOS = 0, and VCOCAL[1:0] = 10 binary by default.

LOS=1?

Y

LOLonLOS = 0 and
ltrOnLos=0? (Default)

Y

CDR continuously tries to lock to
incoming data, and the VCO
frequency does not get re-centered
prior to the lock acquisition
process.

LOLonLOS =1
and ltrOnLos =0?

Y

LOLonLOS = 0
and ltrOnLos = 1?

LOLonLOS = 1
and ltrOnLos = 1?

Next Figure

VCOCAL[1:0] = 01?

Y

VCOCAL[1:0] = 00?

Y

Is refClk
present?

Y

N

The internal VCO pull
range will be automatically
re-centered to the
reference clock frequency
to start the CDR lock
acquisition process.

The entire VCO frequency
range will be swiped to start the
CDR lock acquisition process.

VCOCAL[1:0] = 10
(Default)?

Y

Is refClk
present?

N

VCOCAL[1:0] = 11?

Y

VCO stays at the center of its
frequency range awaiting for
refClk.

Invalid mode!
LOL will stay on.

Figure 12. CDR and VCO Behaviors Upon Declaring LOS (1 of 2)

From Previous Figure

LOLonLOS = 0 and
ltrOnLos = 1?

LOLonLOS =1 and
ltrOnLos =1?

Y

Ltr (bit 1 in Register 7)= 1?

Y

Since ltr has been set to 1
prior to the assertion of
LOS, CMU maintains lock
to the reference and the
CDR lock acquisition
depends on CDRLTDATA
control in Register 7.

N

Is refClk
present?

Y

N

VCO stays at the
center of its
frequency range
awaiting for refClk.

CMU locks to the
reference clock until
LOS is clear. Note that
LOL will be off during
this period.

VCOCAL[1:0] = X1?

Y

Invalid mode!
ltrOnLOS is enabled
but refClk is disabled.

VCOCAL[1:0] = X0?

Y

Is refClk
present?

Y

N

The internal VCO pull range
will be automatically re-
centered to the reference
clock frequency to start the
CDR lock acquisition
process. Note that LOL will
not be cleared.

VCO stays at the center of
its frequency range awaiting
for refClk.

Figure 13. CDR and VCO Behaviors Upon Declaring LOS (2 of 2)
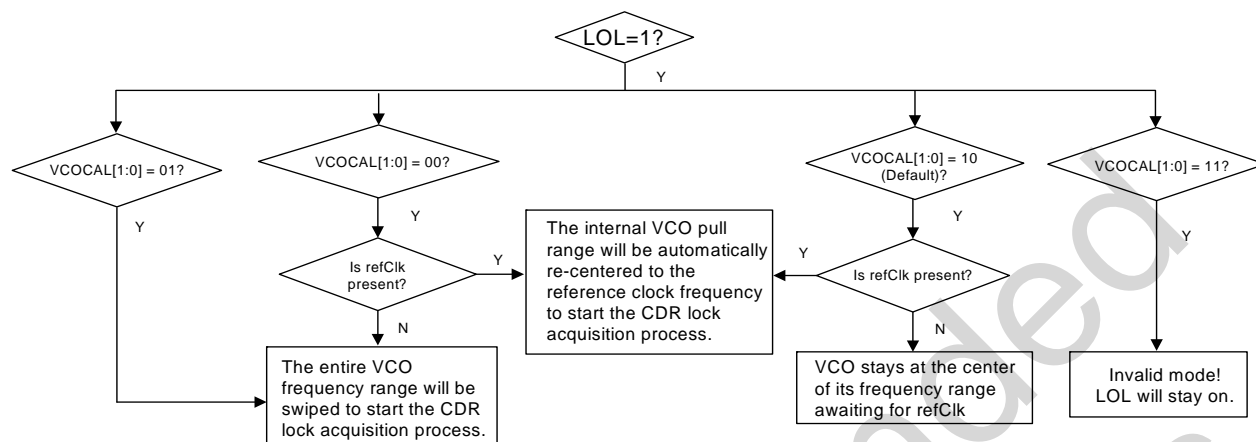
Figure 14. Receive and Transmit CDR and VCO Behaviors Upon Declaring LOL

5.5. Receiver Slice Control

In order to optimize the bit error rate performance of the system, the receiver supports automatic and manual adjustment of the 0/1 decision threshold (slice control). Four slice modes can be programmed via Register 20 as detailed below:

- AutoSlice Mode (sliceEn[2:0] = 001): The slice is automatically set without any control by the user. An internal estimate of the eye opening is used to automatically adjust the slice voltage. This mode can be used when there is no FEC processor as it continually adjusts the slice voltage. Registers 22/21 have no effect on this mode of operation.

- Constant Duty Cycle Mode (sliceEn[2:0] = 010):
  This mode continually adjusts the slice voltage to maintain a user-selected duty cycle at the limiting amplifier output. The duty cycle can be set in the sliceLVL register (Registers 22 and 21). This mode can be used when there is no FEC processor, however, if the duty cycle is dynamically controlled the recommended duty cycle step size is less than or equal to 0.1%. In an actual system the duty cycle should not be set to less than 45% or greater than 55%. For the case of maximum dispersion, an optimally-set duty cycle will typically perform slightly better than Autoslice.

- Proportional Mode (sliceEn[2:0] = 011):The slice offset is defined as a percentage of the peak-to-peak value of the input signal. The percent value is written in the sliceLVL register (Registers 22 and 21) as an offset from 50%. Due to drifts and temperature variations in the silicon, the slice offset values must be dynamically modified at a rate of 100 ms or faster. This mode is recommended only when a FEC processor is present to control the slice level.

- Absolute Mode (sliceEn[2:0] = 100): The slice offset is defined as an absolute voltage. The offset can be set in the range of –240 to +240 mV in the sliceLVL register (Registers 22 and 21). Just as in proportional mode, the slice offset values must be dynamically modified at a rate of 100ms or faster. This mode is recommended only when a FEC processor is present to control the slice level.

Autoslice and constant duty cycle are the preferred slice modes of operation for Telecom and Datacom applications.

5.6. Clock and Data Recovery (CDR)

The Si5040 integrates a CDR to recover the clock and data from the input signal applied to RXDIN. The CDR can be operated with or without an external reference clock. Reference or referenceless operation is programmed in the RxCalConfig register (Register 8). If a reference clock is applied to the receiver, the CDR can be forced to lock to the reference clock in the event that a loss of signal occurs. The CDR can be programmed to continue to sample the RXDIN input while the receive PLL is locked to the reference clock. These options are controlled in the RxConfig register (Register 7).

5.7. Reference Clock

The Si5040 will operate with or without an external reference clock. If a reference clock is applied, the receiver uses the reference clock to center the internal VCO pull range, which, in turn, reduces the acquisition time of the CDR. If the reference clock is not applied, the entire VCO frequency range will be swept for lock acquisition.

Note that since the applied reference clock is used for both the receiver and the transmitter and since the receiver may be running at a different rate than the transmitter, the user is given the option of disabling the reference clock on the receiver.

The RxrefclkEn bit in the RxConfig register (Register 7) controls this function. The receiver can be locked to the reference clock under the following programmable conditions: (RxConfig, Register 7)

1. Set LTR (bit 1).
2. Set LTR on receive loss-of-signal (LOS) (bit 5).

Note: If it is desired to allow the CDR to acquire lock to the incoming data while LTR at Register 7, Bit 1 is set to 1 (Lock to Reference clock enabled), set CDRLTDATA at Register 7, Bit 4 to 1 (default). If it is desired to sample the incoming data with a programmable phase and slice level while LTR is set to 1, set sliceEn[2:0] at Register 20 to 000 binary (auto slice disabled).

5.8. Receiver Loss of Lock (LOL)

Receiver LOL functions differently depending on whether the receiver is operating in reference or referenceless mode. By default (uselolmode Register 7, Bit 3 = 0), SQM-based LOL is used in referenceless mode, and Frequency-based LOL is used in reference mode. In reference mode however, either SQM or Frequency LOL can be used by setting Register 7, Bits 2 and 3 to the appropriate values.

5.8.1. SQM LOL

SQMLOL mode is selected in one of two ways. If register 7[3:2] = 11b, then SQMLOL mode is selected. If register 7[3] = 0 and register 7[1] = 0, then SQMLOL mode is selected. The SQMLOL method compares an internal jitter measure to the sqmLOLThresh (see below on how to set this threshold). When the internal jitter measure is greater than the sqmLOLThresh, RXLOL is asserted. When RxLOL is asserted the 5040 RX side will automatically start to try to acquire lock again across an input data range of 9.8–11.4 Gbps. RxLOL is deasserted when the jitter measure is less than the sqmLOLThresh. The sqmLOLThreshold must be set using registers 106, 107, 108, and 109 in the following order:

1. Write register 107 = A0h.
2. Write register 108 = 3Fh.
3. Write register 109 = B9h.
4. Write register 106 = 04h.
5. Write register 106 = 84h.

These are indexed address registers. Register 106 contains the sqmLOLThresh register address and 107-109 contain the data to be written to it. Register 106 must be written twice. The first write of 04h sets the address, and 84h applies the value in 107-109 into the sqmLOLThresh registers. The above values are recommended for all applications.

Using sqmLOLThresh values other than the default or the one given above can cause unexpected problems, such as false lock, and are not recommended.

5.8.1.1. Dynamic Register Control

The dynamic control of RxLoopFAcq (Register 98) is required to ensure the locking performance of the CDR. It is required for all applications that RxLoopFAcq be set to 98h when RX LOL is asserted and to 00h when RX LOL is deasserted. Only the default value and the value given above are supported for writes to Register 98. Any read back of this register will not necessarily return the value written. If a valid reference clock is applied at pins 13,14 and rxRefclkEn = 1 (reg7[0]) and Rx VCOCAL = x0b (reg8[2:1]), then the dynamic register write to register 98 is not necessary.

In addition, for proper LOL performance, RxPDGainAcq (Register 77) must be written once to 0Dh after power is applied or a SW reset is implemented. If a valid reference clock is applied at pins 13,14 and rxRefclkEn = 1 (reg7[0]) and Rx VCOCAL = x0b (reg8[2:1]), then it is not necessary to write to register 77.

5.8.2. Frequency LOL

The Si5040 supports the use of a ~622 MHz or ~155 MHz (/64 or /16) reference clock. The reference clock frequency is selected in the ChipConfig1 register (Register 2). There are two ways in which FREQLOL is selected. When register7[3:2] = 10b, then FREQLOL is selected. When register7[3] = 0 and register7[1] = 1, then FREQLOL is selected. LOL is asserted if the recovered clock frequency deviates from the reference clock frequency by >±1000 ppm. LOL is de-asserted if the recovered clock is within ±200 ppm of the reference clock frequency. Refer to Figure 14 for CDR and VCO behaviors upon declaring LOL.

5.8.3. Acquisition Time Enhancement

The acquisition lock time for a signal applied at RXDIN can be reduced to less than 15 ms by the following register writes:

1. Write register 86 = 0011 1000 = 38h.
2. Write register 67 = 0100 0001 = 41h.
3. Write register 68 = 0000 0011 = 03h.

5.8.4. LOL Interrupt

LOL may be configured to generate an interrupt. The status of the LOL interrupt bit can be read from the RxintStatus register (Register 5). The status of LOL may also be read from the RxAlarmStatus register (Register 9). LOL may also be asserted upon activation of LOS (see "5.4. Receiver Loss of Signal Alarm (LOS)" on page 20 and Figure 15 on page 26). Receive data (RD) may be squelched on LOL. This option is configured in the RxdPathConfig register (Register 28).

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aLOSThresh (Bit 1:0, Register 13 and Bit 7:0, Register 12)
aLOSHyst[3:0] (Bit 7:4, Register 13)

Peak-to-Peak Monitor

aLOS
(Bit 1, Register 11)

aLosEn
(Bit 0, Register 10, Default= 1)

RxdLosAssertThresh[7:0] (Bit 7:0, Register 17)
RxdLosClearThresh (Bit 7:0, Register 18)

DLOS Monitor
EN

dLosEn[1:0]
(Bit 2:1, Register 10)
00: Disabled (Default)
01: Based on consecutive number of 1s
01: Based on consecutive number of 0s
01: Based on either consecutive number of 1s or 0s

dLOS
(Bit 2, Register 11)

LOS
(Bit 5, Register 9,
Bit 0, Register 11 or Pin 3)

RxSqmThresh[5:0] (Bit 7:2, Register 26)
RxSqmDeassertThresh[5:0] (Bit 5:0, Register 27)

Signal Quality Monitor
EN

1

sqmAlarm
(Bit 0, Register 9)

sqmLOS
(Bit 4, Register 11)

sqmLosEn
(Bit 3, Register 10, Default= 0)

RxSqmValue[5:0]
(Bit 5:0, Register 25)

lolOnLos
(Bit 6, Register 7, Default= 0)

SqmLol

1 (Default)

Rx Recovered Clock

Frequency Offset Monitor

FreqLol

Reference Clock

0
Select

LOL
(Bit 4, Register 9 or Pin 2)

lolMode
(Bit 2, Register 7)

useLolMod
(Bit 3, Register 7, Default= 0)

Figure 15. RX LOS and LOL Block Diagram

TxdLosAssertThresh[7:0] (Bit 7:0, Register 145)
TxdLosClearThresh (Bit 7:0, Register 146)

DLOS Monitor

EN

dLosEn[1:0]
(Bit 2:1, Register 138)
00: Disabled (Default)
01: Based on consecutive number of 1s
01: Based on consecutive number of 0s
01: Based on either consecutive number of 1s or 0s

dLOS
(Bit 2, Register 139)

LOS
(Bit 5, Register 137 or
Bit 0, Register 139)

TxSqmThresh[5:0] (Bit 7:2, Register 154)
TxSqmDeassertThresh[5:0] (Bit 5:0, Register 155)

Signal
Quality
Monitor

1 ——▶ EN

sqmAlarm
(Bit 0, Register 137)

sqmLosEn
(Bit 3, Register 138, Default= 0)

sqmLOS
(Bit 4, Register 139)

TxSqmValue[5:0]
(Bit 5:0, Register 153)

SqmLol

1 (Default)

Tx Recovered Clock ——▶

Frequency
Offset
Monitor

Reference Clock ——▶

FreqLol

0

Select

LOL
(Bit 4, Register 137)

lolMode
(Bit 2, Register 135)

useLolMod
(Bit 3, Register 135, Default= 0)

Figure 16. TX LOS and LOL Block Diagram

5.9. Receiver Phase Adjust

The Si5040 receiver supports programmable sample phase adjust. The sampling point may be advanced or delayed in time by adjusting the value loaded into the RxPhaseAdjust register (Register 24). The range of adjustment is ±12 ps.

5.10. Receive Clock Multiplier Unit

The Si5040 receiver incorporates a DSPLL®-based clock multiplier unit (CMU) that attenuates the jitter on data recovered from the line interface. This makes it much easier to significantly exceed the jitter requirements for 10 Gbit SONET, Ethernet, and Fibre Channel applications. The CMU is rate-adaptable across the entire range of device operation. Note that when the ltr bit or ltrOnLOS bit in Register 7 is set to 1, the receive CMU is locked to the reference clock.

The receiver CMU supports 380 kHz bandwidth (cmuBandwidth[3:0] at Register 6 = 0100).

5.11. Recommended Pre-Emphasis on the RD Signal

Even though the RD signal rise/fall time is very fast, some users may wish to add high-frequency boost to the RD signal. This is done with an external RC network to reduce low-frequency energy, effectively boosting the high frequencies. To compensate for loss in the circuit and maintain proper signal size and eye opening at the XFI connector, the RD amplitude will have to increase to 700 or 800 mV. Register 56 controls the RD signal amplitude. The resistors and capacitors can be generic low-cost components, and the circuit should be located very close to the Si5040 RD± pins. This circuit is recommended for all XFP applications.
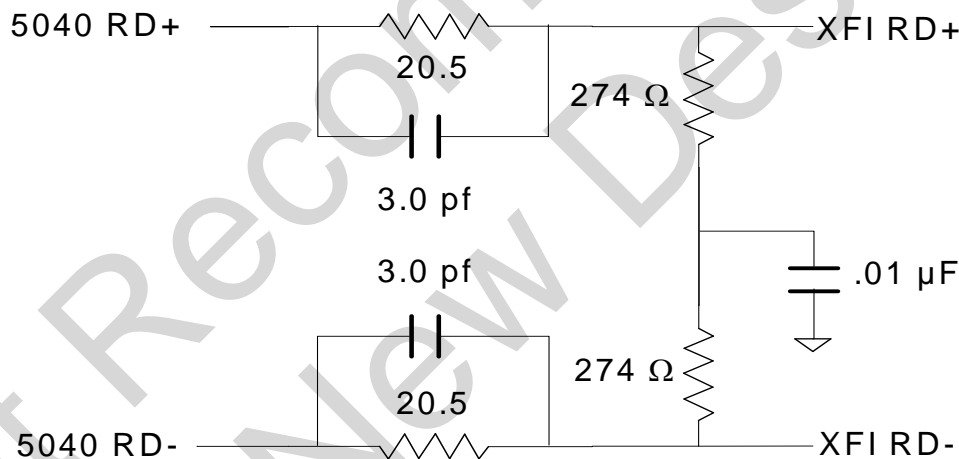
Figure 17. RD Pre-Emphasis Circuit

6. Transmitter

The Si5040 transmitter includes an XFI-compliant, fixed-equalizer CDR for recovery of clock and data from the XFI channel (TD inputs), pattern generation and checking function (see "6.3. Clock and Data Recovery (CDR)"), transmit FIFO, and jitter-attenuating clock multiplier unit.

6.1. Transmitter Loss-of-Signal Alarm (LOS)

The Si5040 transmitter generates a loss-of-signal alarm when the TD input signal fails to meet the selected programmable condition for Transmit Loss of Signal. The programmable LOS mode is controlled in the TxLosCtrl register (Register 138). The available modes are Digital Loss of Signal (DLOS) and Signal Quality Monitor (SQM). The state of LOS is reflected in the LOS bit in the TxLosStatus register (Register 139). LOS may also be configured to generate an interrupt. The status of the LOS interrupt bit may be read in the TxintStatus register (Register 133). The status of the various LOS modes is stored in the TxLosStatus register (Register 139).

A DLOS alarm occurs when the bit stream on the TD input contains a run length of ones or zeroes greater than the value loaded in the TxdLosAssertThresh register (Register 145). dLos will remain asserted until the bit stream shows activity for a time greater than that loaded in the TxdLOSClearThresh register (Register 146).

An SQM alarm occurs when the TD input signal quality falls below the value loaded in TxSqmThresh (Register 154). The Signal Quality Monitor measures the magnitude of the horizontal eye opening of the received signal. The SQM value can be read from the TxsqmValue register (Register 153). An SQM alarm will assert if the TxsqmEn bit has been set in the TxSqmConfig register (Register 154). SQM hysteresis is set in the TxsqmDeassertThresh register (Register 155).

The transmitter may be programmed to cause the following events on an LOS condition:

1. Disable (squelch) the transmitter data output (TXDOUT) (Register 156).
2. Generate a clock pattern at the transmit data output (TXDOUT) (Register 156).

6.2. Transmit Equalizer

The TX equalizer is a passive, fixed-gain equalizer based on the inverse response of the XFI channel. The equalizer attenuates the low-frequency components of the input signal to compensate for the high-frequency losses through the XFI channel. The overall frequency response through the XFI channel and the equalizer should be essentially flat.

6.3. Clock and Data Recovery (CDR)

The Si5040 integrates a CDR to recover the clock and data from the signal applied to the TD input. The CDR may be operated with or without an external reference clock. Reference and referenceless operation is programmed in the TxCalConfig register (Register 136). If a reference clock is applied, the CDR may be forced to lock to the reference clock in the event that a loss of signal occurs. This option is programmed in the TxConfig register (Register 135).

6.4. Transmitter Loss of Lock (LOL)

Transmitter LOL functions in different ways depending on whether the transmitter is operating in reference or referenceless mode. By default (uselolmode Register 135, Bit 3 = 0), SQM-based LOL is used in referenceless mode, and Frequency-based LOL is used in reference mode. However, in reference mode, either SQM or Frequency LOL can be used by setting Register 135, Bits 2 and 3, to the appropriate values.

6.4.1. SQM LOL

When the VCO is configured to calibrate without a reference clock (VCOCAL[1:0] = 01 binary), the default values of register 135[3:2] will cause the LOL method to be SQMLOL. Just as in the receiver (see "5.8.1. SQM LOL" on page 24), the SQMLOL method compares an internal jitter measure to the sqmLOLThresh; however, the TX sqmLOLThresh value must not be modified. When the internal jitter measure is greater than the sqmLOLThresh, TXLOL is asserted. TxLOL is deasserted when the jitter measure is less than the sqmLOLThresh.

6.4.1.1. Dynamic Register Control

The dynamic control of TxLoopFAcq (Register 226) is required to ensure the locking performance of the CDR. For all applications, it is required that TxLoopFAcq be set to 98h when TX LOL is asserted and to 00h when TX LOL is deasserted. Only the default value and the value given above are supported for Register 226. Any read back of this register will not necessarily return 98h. If a valid reference clock is applied at pins 13,14 and the Tx VCOCAL = x0b (reg136[2:1]), then the dynamic register write to register 226 is not necessary.

In addition, for proper LOL performance, TxPDGainAcq (Register 205) must be written once to 0Dh after power is applied or a SW reset is implemented. If a valid reference clock is applied at pins 13,14 and Tx VCOCAL = x0b (reg136[2:1]) then it is not necessary to write to register 205.

6.4.2. Frequency LOL

The Si5040 supports the use of a ~622 MHz or ~155 MHz (/64 or /16) reference clock. When FREQLOL is set (Register 135[3:2] = 10b), LOL is asserted if the recovered clock frequency deviates from the reference clock frequency by >±1000 ppm. LOL is de-asserted if the recovered clock is within ±200 ppm of the reference clock frequency. The reference clock frequency is selected in the ChipConfig1 register (Register 2). Refer to Figure 14 for CDR and VCO behaviors upon declaring LOL.

6.4.3. Acquisition Time Enhancement

The acquisition time for a signal applied at TD can be reduced to less than 15 ms by the following register writes:

- Register 195 = 0100 0000 = 40h
- Register 196 = 0000 0111 = 07h
- Register 214 = 0011 1000 = 38h

6.4.4. LOL Interrupt

LOL may be configured to generate an interrupt. The status of the LOL interrupt bit can be read from the TxintStatus register (Register 133). The status of LOL may also be read from the TxAlarmStatus register (Register 137). Transmitter data (TXDOUT) may be squelched on LOL. This option is configured in the TxdPathConfig register (Register 156).

6.5. Transmitter Phase Adjust

The Si5040 transmitter supports manual sample phase adjust. The sampling point may be advanced or delayed in time by adjusting the value loaded into the PhaseAdjust register (Register 152). The range of adjustment is >±12 ps. Note that the transfer function from the register value to the phase adjust time is highly variable; therefore, we only guarantee that the largest or smallest register value will achieve better than +12 ps or –12 ps, respectively.

6.6. Transmit Clock Multiplier Unit

The Si5040 transmitter incorporates a DSPLL®-based clock multiplier unit (CMU) that attenuates the jitter on the data recovered from the XFI interface. This makes it much easier to significantly exceed the jitter requirements for 10 Gbit SONET, Ethernet, and FibreChannel applications. The CMU is rate-adaptable across the entire range of device operation. Selectable CMU bandwidths support adjustment of the degree of jitter filtering required for a given application.

6.7. Timing Modes Of Operation

For maximum flexibility, the Si5040 supports three CMU timing modes that make it suitable for XFP modules targeted at both datacom and telecom applications. The modes of operation determine how the transmit CMU is configured. Timing modes are set in the TxCmuConfig register (Register 134).

6.7.1. Referenceless Mode (Mode 0)

In the referenceless mode of operation, timing information is recovered from the XFI data and used as the timing source for the transmit CMU (Figure 18). The mode is set in TxCmuConfig (Register 134). This mode should be chosen when a synchronous reference clock is not available or if the jitter on the recovered XFI data is less than that on the available synchronous reference clock.

Optionally, an asynchronous reference clock at a frequency of 1/64 of the baud rate (as defined in the XFP specification) may be applied for VCO centering purposes and to generate receive LOL and transmit LOL. However, a reference clock is not necessary to generate these signals. Referenceless mode is the default mode after power-on.

6.7.2. Synchronous Reference Clock Mode (Mode 1)

If an external synchronous reference clock with jitter characteristics of higher quality than those of the recovered XFI clock is available, the device may be operated in Synchronous Reference Clock Mode (Figure 19). The mode is set in TxCmuConfig (Register 134). In this mode, the transmit CMU derives the line-rate clock by multiplying the clock frequency applied to the REFCLK inputs by 64. This mode is equivalent to the "Optional Synchronous CMU Clock" mode described in the XFP specification (see XFP MSA Rev. 4.0 Section 3.9.1). It is not necessary to meet the phase noise characteristics defined for the synchronous reference clock in the XFP specification since the Si5040 transmit CMU attenuates jitter on the reference clock.

6.7.3. Asynchronous Reference Clock Mode (Mode 2)

If an external asynchronous reference clock with jitter characteristics of higher quality than those of the recovered XFI clock is available, the device can be operated in Asynchronous Reference Clock Mode (Figure 20). The mode is set in TxCmuConfig (Register 134). The external reference clock is used as the timing source for the transmit CMU. The resulting transmit clock is frequency locked to the incoming data, and the jitter on the transmit clock is reduced due to the low jitter of the applied reference clock. The transmit CMU bandwidth may be set to 180 or 1370 Hz, depending on how much jitter is present on the applied reference clock. A FIFO in the data path accommodates any jitter differences between the serial data and the CMU line-rate clock.

Figure 18. Referenceless Mode (Mode 0)

Figure 19. Synchronous Reference Clock (Mode 1)
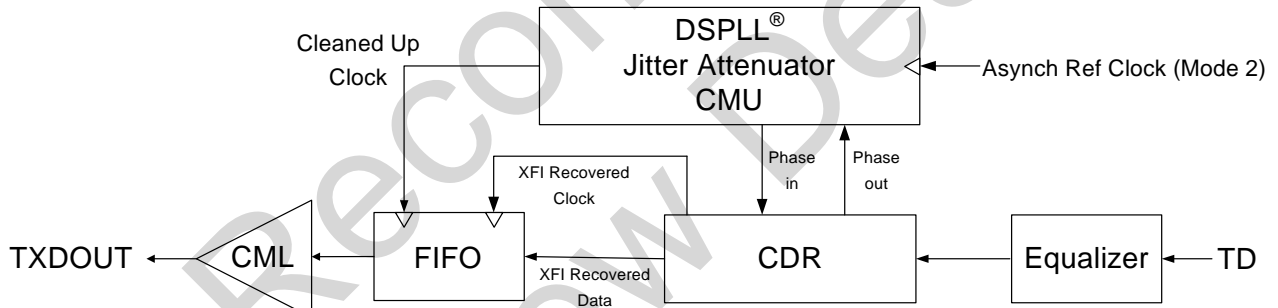
Figure 20. Mode 2

7. Loopback Modes

The Si5040 supports XFI Loopback, Lineside Loopback, and Looptime modes.

7.1. XFI Loopback

The Si5040 is configured in the XFI Loopback mode by writing to the ChipConfig1 register (Register 2). The Si5040 is configured in the XFI Loopback mode by writing to the ChipConfig1 register (Register 2). Data on the TD input is retimed and output on the RD output. The clock recovered from the XFI data (TD) is used as the timing source for the RD output. For this reason, RX VCOCAL (Register 8, Bit [2:1]) must be set to reference or auto mode. If the data from the transmitter is not required at the TXDOUT output pins, it may be disabled by writing to the transmitter Squelch bit in the dPath register (Register 156, bit 2). Data on the TD input is retimed and output on the RD output. The clock recovered from the XFI data (TD) is used as the timing source for the RD output.

7.2. Lineside Loopback

The Si5040 is configured in the Lineside Loopback mode by writing to the ChipConfig1 register (Register 2). Data received on the receiver input (RXDIN) is output on the transmitter (TXDOUT) output. Since in this mode the clock recovered from the receiver input is used as the timing source for the transmit CMU, TX VCOCAL (Register 136, Bit [2:1]) must be in reference or auto mode. If the data from the receiver is not required at the XFI interface, it may be disabled by writing to the receiver Squelch bit in the dPath register (Register 28, bit 2).

8. Looptime Mode

The Si5040 supports looptime mode for applications in which it is desirable to time the transmitter off the receiver clock. Data received at the XFI interface (TD) is retimed using the clock recovered from the receiver (RXDIN). As a result of the transmit CMU jitter attenuation feature in the Si5040, with the appropriate setting of the CMU bandwidth (set in Register 134), the jitter on the recovered clock is significantly attenuated so that the data on the transmit output will be compliant with the datacom and telecom standards supported by the Si5040. A FIFO within the data path accommodates any jitter differences between the serial data and the CMU line-rate clock. Looptime mode is enabled by writing to the TxCmuConfig register (Register 134).
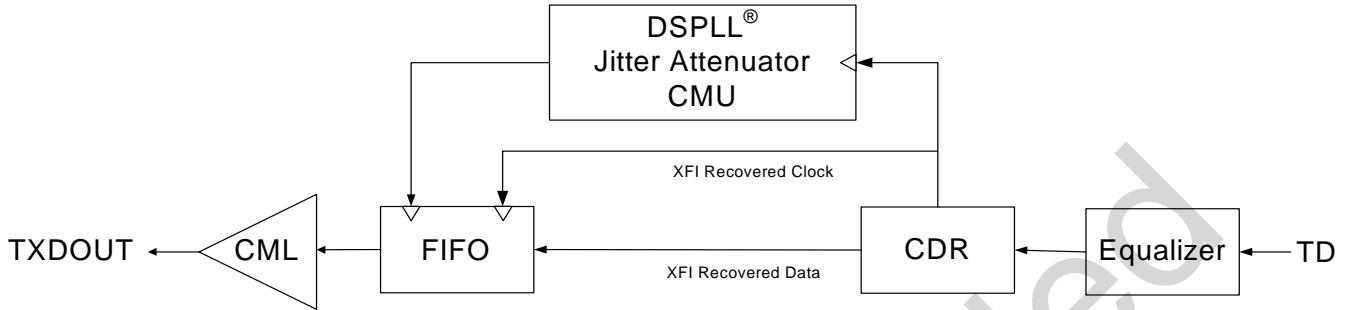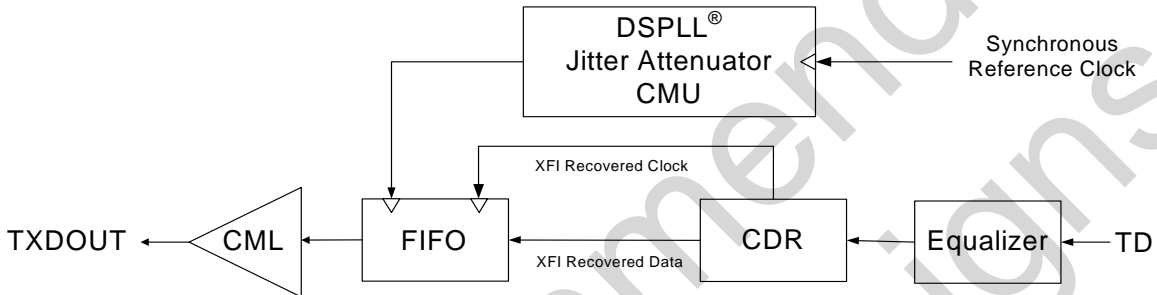
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9. Pattern Generation and Checking

The Si5040 includes a programmable pattern generator and checker function in both the receiver and transmitter signal paths. The Si5040 can generate and check PRBS7, PRBS31, or a 64-bit, user-defined pattern programmed in the tpSel register (receiver Register 29, transmitter Register 157).

Notes:
1. When PRBS7 or PRBS31 is selected for the pattern generator or checker, the pattern can either be inverted or non-inverted by programming the tpGenInvert and tpChkInvert bits in the pgSel register (receiver Register 29, transmitter Register 157). Per Section 5.8 in O.150, PRBS31 is specified as an inverted pattern. The pattern generators default to generating an inverted PRBS31 to comply with 0.150. However, the pattern generators and checkers have the option to invert the pattern.
2. The pattern checker will report no error if the input sequence is an all 0s pattern. However, a loss-of-lock or loss-of-signal indicator will assert in these conditions.

The user-defined patterns are programmed in the tpArbGenPtn and tpArbChkPtn registers (receiver Registers 31–38 and transmitter Registers 159–166). The time period or number of bits over which the checker should look for errors is defined in the tpTimeBase bits located in the tpChkConfig register (receiver Register 30, transmitter Register 158). The time base can be programmed to be infinite (always looking for errors) or set to one of three defined values. Changing to another time base will reset the error counter.

The pattern checker offers a Loss-of-Sync indicator along with a 40-bit error count register and an 8-bit error count register in floating point notation. When the checker achieves synchronization between the expected and the received pattern, the tpSyncLos register (receiver Register 9, transmitter Register 137) is deasserted. Note that as soon as the checker is synchronized, the error count register is reset to 0, and error counting begins. As soon as the checker loses synchronization in the middle of a measuring window defined by the tpTimeBase register, the error count register is loaded with all 1s, indicating the maximum error count. In order to differentiate between a Loss-of-Sync event and a bit error event, the user should monitor both the tpSyncLos register and the error count register described below.

The tpChkErrCnt register holds the error count from the last completed time base if the checker is in a defined time base mode; otherwise, in the infinite time base setting, the current running error count is stored. An interrupt is generated when the number of errors exceeds the value loaded in the 8-bit tpTargetErr register (receiver Register 47, transmitter Register 175). The tpChkErr register holds the error count in an 8-bit floating point format.

In order for the pattern generation/check function to operate correctly, a timing source must be applied. A valid timing source for the receiver pattern generation/check function can be any of the following:

- A reference clock with the device placed in lock-to-reference mode.
- Data applied at the RXDIN inputs from which a recovered clock can be derived.
- Data applied at the TD inputs from which a recovered clock can be derived and with the device placed in XFI loopback mode.

A valid timing source for the transmitter pattern generation/check function can be any of the following:

- A reference clock with the device placed in lock-to-reference mode.
- Data applied at the TD inputs from which a recovered clock can be derived.
- Data applied at the RXDIN inputs from which a recovered clock can be derived and with the device placed in Looptime mode.

10. Serial Microcontroller Interface

Device control and status monitoring is supported with a selectable I^{2}C or SPI-like interface. SPSEL (Pin 9) controls which of the two serial formats is selected.

10.1. I^{2}C Interface

When configured in I^{2}C control mode (pin SPSEL tied low), the control interface to the Si5040 is a 2-wire bus for bidirectional communication. The bus consists of a bidirectional serial data line (SD) and a serial clock input (SCK). The SD pin may be configured as a CMOS output or as an open drain output using Register 2, bit 4. Fast mode operation is supported for transfer rates up to 400 kbps as specified in the I^{2}C-Bus Specification standard. A chip select pin is provided (SS) to address the Si5040.

Figure 21 shows the command format for both read and write access. Data is always sent MSB first. Table 2 on page 6 and Table 3 on page 8 give the dc and ac electrical parameters for the SCK and SD I/Os, respectively. The timing specifications and timing diagram for the I^{2}C bus can be found in the I^{2}C-Bus Specification standard (fast mode operation).

| S | Slave Address | 0 | A | Byte Address | A | Data | A | Data | A | P |

Write Command

| S | Slave Address | 0 | A | Byte Address | A | S | Slave Address | 1 | A | Data | A | Data | Ā | P |

Read Command

Address auto incremented after each data read or write

1 Read   0 Write

From master to slave   A – Acknowledge (SD LOW)
S – START condition
From slave to master   P – STOP condition
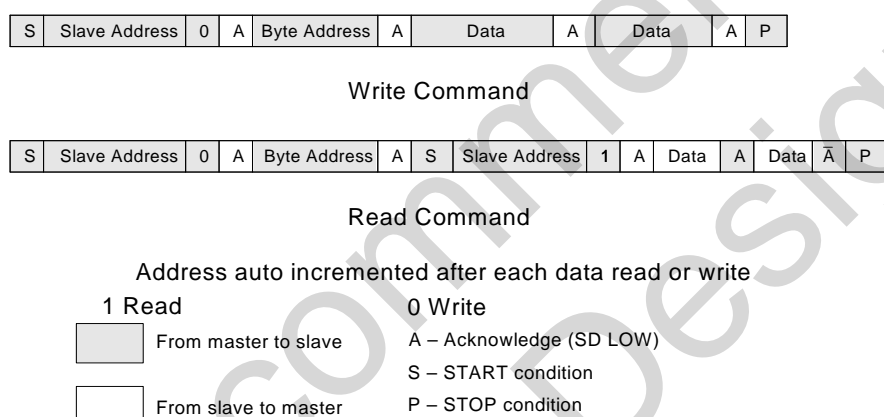
Figure 21. I^{2}C Command Format

The device has two possible I^{2}C addresses depending on the SSb pin setting. If the SSb pin is floating or externally tied high, the device has an I^{2}C address of 7d' 1000001. If the SSb pin is externally tied low, the device has an I^{2}C address of 7d' 1000000. For applications that require two Si5040 devices connected on the same I^{2}C bus, each individual device can be accessed with a unique I^{2}C address depending on the SSb pin setting.

Figure 22 illustrates how the I^{2}C address can be configured and what the expected value is for the first byte after the START condition. Note that the first byte after the START condition could be either 82h or 83h depending on whether it's a read or write. The first byte after the START condition could be either 80h or 81h depending on whether it is a read or write.

By design, this bit in the I^{2}C address is determined by the SSb pin.   Per I^{2}C specification, this bit is either 1 for read or 0 for write.

MSB   LSB
1 0 0 0 0 0 X R/W
slave address   MBC608

The first byte after the START procedure

Source: Fig 14 of I^{2}C – Bus Specification Version 2.1
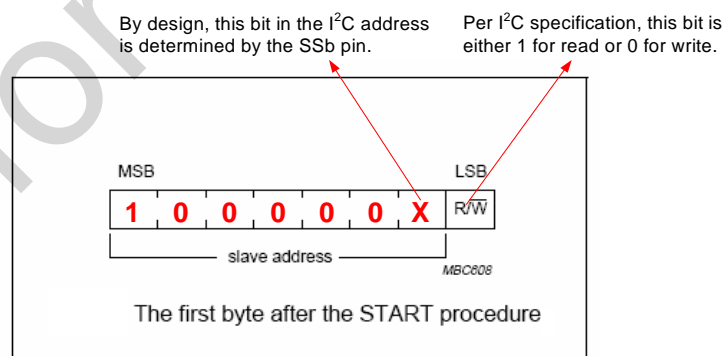
Figure 22. Device I^{2}C Address

10.2. SPI-Like Interface

When configured in SPI-like control mode (pin SPSEL tied high), the control interface to the Si5040 is a 3-wire interface modeled close to commonly-available microcontrollers and bidirectional serial peripheral devices. The interface consists of a clock input (SCK), slave select input (SS), and serial data input/output (SD). The SD pin may be configured as a CMOS output or as an open drain output using Register 2, bit 4.

Data is transferred one byte at a time, with each register access consisting of a pair of byte transfers. Figure 7 and Figure 8 on page 15 illustrate read and write/set address operations on the SPI bus, and Table 9 on page 14 gives the timing requirements for the interface. Table 12 shows the SPI command format.

The first byte of the pair is the instruction byte. The "Set Address" command writes the 8-bit address value that will be used for the subsequent read or write.

The "Write" command writes data into the device based on the address previously established, and the "Write/Address Increment" command writes data into the device and automatically increments the register address for use on the subsequent command. The "Read" command reads one byte of data from the device, and the "Read/Address Increment" reads one byte and increments the register address automatically.

The second byte of the pair is the address or data byte. As shown in Figure 7 and Figure 8 on page 15, SS should be held low during the entire two byte transfer. Raising SS resets the internal state machine; so, SS must be raised between each two byte transfers to guarantee that the state machine will be reinitialized.

During a read operation, the SD becomes active on the falling edge of SCK, and the 8-bit contents of the register are driven out MSB first. The SD is high-impedance on the rising edge of SS. During write operations, data is driven into the Si5040 via the SD pin MSB first. Data always transitions with the falling edge of the clock and is latched on the rising edge.

The clock should return to a logic high when no transfer is in progress. The Si5040 SPI-like interface supports continuous clocking operation where SS is used to gate two byte transfers.

Table 12. SPI-Like Command Format

Instruction	Address/Dat
00000000—Set Address	AAAAAAAA
01000000—Write	DDDDDDDD
01100000—Write/Address Increment	DDDDDDDD
10000000—Read	DDDDDDDD
10100000—Read/Address Increment	DDDDDDDD

11. Interrupt Functionality

Alarm Status bits (Register 9/137) are sampled by a 10 MHz clock to create the Interrupt Status bits (Register 5/133). If the Interrupt Enable bit (Register 2, bit 5) is a zero, all the Interrupt status bits are forced to zero. The Alarm Status bits are always active regardless of the state of the Interrupt Enable bit and the Interrupt Mask bits. The Interrupt Mask bits (Register 4/132) masks the Alarms Status bits from affecting the corresponding Interrupt Status bit. Writing a zero to an Interrupt Status bit forces that bit to a zero; however, if the corresponding Alarm bit is active and the Interrupt Mask bit is not set, the next 10 MHz clock cycle will again cause the Interrupt Status bit to set. All of the Interrupt Status bits are logically "NORed" together to create the Interrupt bit (or pin16). Please refer to Figure 23 for an illustration of the device interrupt tree. The polarity of the Interrupt (pin 16) is active low.

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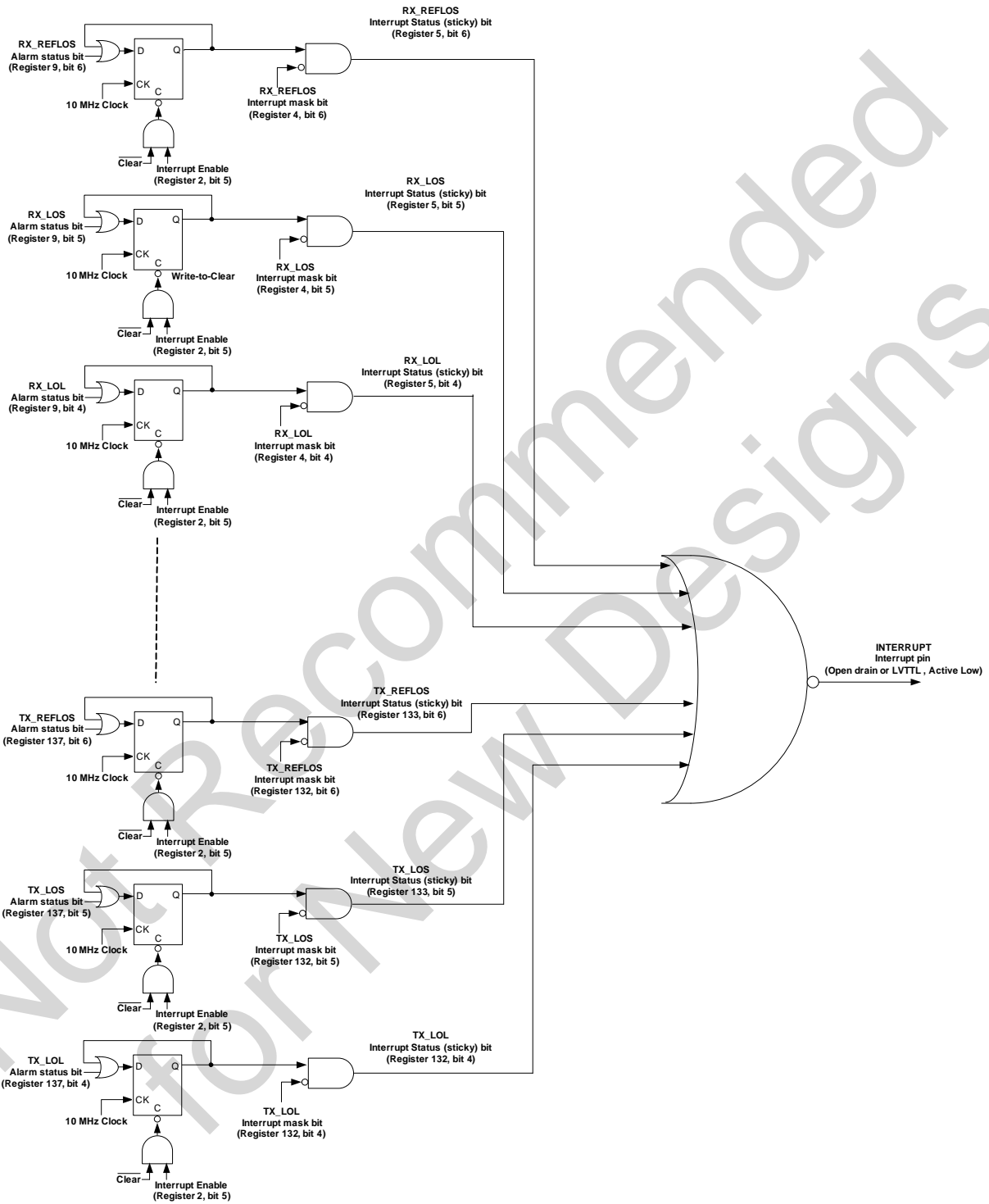
Figure 23. Device Interrupt Tree

12. Programmable Power-Down Options

The RX and TX paths can be powered down independently by programming RxPdn = 1 at Register 3, Bit 0 (Default = 0) or TxPdn = 1 at Register 131, Bit 0 (Default = 0), respectively. As long as both paths are not powered down, all registers are still accessible. Any values written to the registers of the powered-down path will have no immediate effect but will take effect once the path is powered back up and a recalibration is initiated (RX hardRecal at Register 8, Bit 3 or TX hardRecal at Register 136, Bit 3). Any read of registers of the powered-down path will return garbage as there is no VCO clock.

When both the RX and TX paths are powered down, there is a special power up requirement that depends upon the state of the SPSEL pin. If SPSEL is high, any read or write access will wake up the device. The path that was last powered down will be powered up first. The other path can then be powered up by setting the appropriate Pdn bit to a zero. Each side that is powered up must have a hard recal performed to calibrate all circuits (RX hardRecal at Register 8, Bit 3, or TX hardRecal at Register 136, Bit 3).

If SPSEL is low (I^2C mode), power up from both sides powered down can only be accomplished by physically removing VDD from the Si5040 and then re-applying VDD. Of course, this will power up both sides of the device, and a hard recal will automatically occur.

Note that interrupts must be masked or ignored during power down and unmasked and cleared after the recalibration.

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13. Si5040 Register Summary

Any reserved bits listed in the table below or reserved registers (23, 54–55, 58–76, 78–83, 86–97, 99–105, 110–130, 140–144, 148–151, 182–183, 185–204, 206–225, and 227–255) must not be written to a non-default value. All reserved bits have the default values shown below.

Reg	Name	Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Receiver										
0	Part Identifier	40h	Identifier[7:0]							
1	Part Identifier	30h	Revision[3:0]				Identifier[3:0]			
2	ChipConfig1	58h	losOpenDrain	intOpenDrain	intEnable	spiOpenDrain	Reserved	linesideLoopback	XFILoopback	refClkFreq
3	RxChipConfig2	0h	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	RxPdn
4	RxintMask	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
5	RxintStatus (Sticky Bits)	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
6	RxCmuConfig	40h	cmuBandwidth[3:0]				Reserved	Reserved		
7	RxConfig	15h	Reserved	lolOnLOS	ltrOnLOS	CDRLTDATA	uselolMode	lolMode	ltr	rxRefclkEn
8	RxCalConfig	0h	Reserved	Reserved	Reserved	Reserved	hardRecal	VCOCAL[1:0]		swReset
9	RxAlarmStatus	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
10	RxLosCtrl	1h	Reserved	Reserved	Reserved	Reserved	sqmLosEn	dLosEn[1:0]		aLosEn
11	RxLosStatus	0h	Reserved	Reserved	Reserved	sqmLos	dLOSlastTrigger	dLOS	aLOS	LOS
12	aLosThresh	Fh	aLOSThresh[7:0]							
13	aLosThresh2	70h	aLosHyst[3:0]				Reserved		aLosThresh[1:0]	
15	peakDet	0h	peakDet[7:0]							
16	peakDet	0h	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	peakDet[1:0]	
17	RxdLosAssertThresh	5h	RxdLosAssertThresh[7:0]							
18	RxdLosClearThresh	60h	RxdLosClearThresh[7:0]							
20	sliceConfig	11h	Reserved	Reserved	Reserved	Reserved	Reserved	sliceEn[2:0]		
21	sliceLvl	0h	sliceLvl[7:0]							
22	sliceLvl	0h	sliceLvl[15:8]							
24	RxphaseAdjust	0h	Reserved	RxphaseAdjust[6:0]						
25	RxSqmValue	N/A	Reserved	Reserved	RxSqmValue[5:0]					
26	RxSqmConfig	29h	RxSqmThresh[5:0]						Reserved	Reserved
27	RxSqmDeassertThresh	Fh	Reserved	Reserved	RxSqmDeassertThresh[5:0]					
28	RxdPathConfig	2h	Reserved	dinvert	clkOnLOS	SquelchOnRxLOL	SquelchOnRxLOS	Squelch	FIFOAutoReset	FIFOReset
29	RxtpSel	0h	tpChkInvert	tpChkSel[2:0]			tpGenInvert	tpGenSel[2:0]		

Reg	Name	Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
30	RxtpChkConfig	2h	Reserved	Reserved	Reserved	Reserved	Reserved	tpSyncMask	tpTimeBase[1:0]	
31	RxtpArbGenPtn	AAh	RxtpArbGenPtn[7:0]							
32	RxtpArbGenPtn	AAh	RxtpArbGenPtn[15:8]							
33	RxtpArbGenPtn	AAh	RXtpArbGenPtn[23:16]							
34	RxtpArbGenPtn	AAh	RxtpArbGenPtn[31:24]							
35	RxtpArbGenPtn	AAh	RxtpArbGenPtn[39:32]							
36	RxtpArbGenPtn	AAh	RxtpArbGenPtn[47:40]							
37	RxtpArbGenPtn	AAh	RxtpArbGenPtn[55:48]							
38	RxtpArbGenPtn	AAh	RxtpArbGenPtn[63:56]							
39	RxtpArbChkPtn	AAh	RxtpArbChkPtn[7:0]							
40	RxtpArbChkPtn	AAh	RxtpArbChkPtn[15:8]							
41	RxtpArbChkPtn	AAh	RxtpArbChkPtn[23:16]							
42	RxtpArbChkPtn	AAh	RxtpArbChkPtn[31:24]							
43	RxtpArbChkPtn	AAh	RxtpArbChkPtn[39:32]							
44	RxtpArbChkPtn	AAh	RxtpArbChkPtn[47:40]							
45	RxtpArbChkPtn	AAh	RxtpArbChkPtn[55:48]							
46	RxtpArbChkPtn	AAh	RxtpArbChkPtn[63:56]							
47	RxtpTargetErr	FFh	RxtpTargetErr[7:0]							
48	RxtpChkErrCnt	N/A	RxtpChkErrCnt[7:0]							
49	RxtpChkErrCnt	N/A	RxtpChkErrCnt[15:8]							
50	RxtpChkErrCnt	N/A	RxtpChkErrCnt[23:16]							
51	RxtpChkErrCnt	N/A	RxtpChkErrCnt[31:24]							
52	RxtpChkErrCnt	N/A	RxtpChkErrCnt[39:32]							
53	RxtpChkErr	N/A	RxtpChkErr[7:0]							
56	OutputLevel	F5h	HsPowerCtl[1:0]		Reserved	Reserved	Reserved	outLevel[2:0]		
77	RxPDGainAcq	8Dh	RxPDGainAcq[2:0]			Reserved	Reserved	Reserved	Reserved	Reserved
84	RxEqConfig1	A1h	RxEqGain			Reserved				
85	RxEqConfig2	E0h	RxEqHFBoost			Reserved				
98	RxLoopFAcq	1Eh	RxLoop FAcqCtl	RxLoopFAcq[6:0]						
106	sqmLOLThreshWrt	00h	sqmLOL ThreshWrt	Reserved	sqmLOLThreshAdd[5:0]					
107	sqmLOLThresh	00h	sqmLOL Thresh[0]	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
108	sqmLOLThresh	00h	sqmLOLThresh[8:1]							
109	sqmLOLThresh	00h	Reserved	Reserved	Reserved	sqmLOLThresh[13:9]				
Transmitter										
131	TxChipConfig2	22h	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	XORATE	TxPdn

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Reg	Name	Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
132	TxintMask	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
133	TxintStatus (Sticky Bits)	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
134	TxCmuConfig	40h	cmuBandwidth[3:0]				Reserved	cmuMode[2:0]		
135	TxConfig	94h	Reserved	Reserved	Reserved	CDRLT-DATA	uselolMode	lolMode	ltr	Reserved
136	TxCalConfig	0h	Reserved	Reserved	Reserved	Reserved	hardRecal	VCOCAL[1:0]		swReset
137	TxAlarmStatus	0h	Reserved	refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
138	TxLosCtrl	Eh	Reserved	Reserved	Reserved	Reserved	sqmLosEn	dLosEn[1:0]		Reserved
139	TxLosStatus	0h	Reserved	Reserved	Reserved	sqmLOS	dLOSlastTrigger	dLOS	Reserved	LOS
145	TxdLosAssertThresh	00	TxdLosAssertThresh[7:0]							
146	TxdLosClearThresh	60h	TxdLosClearThresh[7:0]							
152	TxPhaseAdjust	0h	Reserved	TxphaseAdjust[6:0]						
153	TxSqmValue	0h	Reserved	Reserved	TxSqmValue[5:0]					
154	TxSqmConfig	05h	TxSqmThresh[5:0]						Reserved	Reserved
155	TxSqmDeassertThresh	08h	Reserved	Reserved	TxSqmDeassertThresh[5:0]					
156	TxdPathConfig	2h	Reserved	dinvert	clkOnLOS	SquelchOnTxLOL	SquelchOnTxLOS	Squelch	FIFOAutoReset	FIFOReset
157	TxtpSel	0h	tpChkInvert	tpChkSel[2:0]			tpGenInvert	tpGenSel[2:0]		
158	TxtpChkConfig	2h	Reserved	Reserved	Reserved	Reserved	Reserved	tpSyncMask	tpTimeBase[1:0]	
159	TxtpArbGenPtn	AAh	TxtpArbGenPtn[7:0]							
160	TxtpArbGenPtn	AAh	TxtpArbGenPtn[15:8]							
161	TxtpArbGenPtn	AAh	TXtpArbGenPtn[23:16]							
162	TxtpArbGenPtn	AAh	TxtpArbGenPtn[31:24]							
163	TxtpArbGenPtn	AAh	TxtpArbGenPtn[39:32]							
164	TxtpArbGenPtn	AAh	TxtpArbGenPtn[47:40]							
165	TxtpArbGenPtn	AAh	TxtpArbGenPtn[55:48]							
166	TxtpArbGenPtn	AAh	TxtpArbGenPtn[63:56]							
167	TxtpArbChkPtn	AAh	TxtpArbChkPtn[7:0]							
168	TxtpArbChkPtn	AAh	TxtpArbChkPtn[15:8]							
169	TxtpArbChkPtn	AAh	TxtpArbChkPtn[23:16]							
170	TxtpArbChkPtn	AAh	TxtpArbChkPtn[31:24]							
171	TxtpArbChkPtn	AAh	TxtpArbChkPtn[39:32]							
172	TxtpArbChkPtn	AAh	TxtpArbChkPtn[47:40]							
173	TxtpArbChkPtn	AAh	TxtpArbChkPtn[55:48]							

Reg	Name	Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
174	TxtpArbChkPtn	AAh	TxtpArbChkPtn[63:56]							
175	TxtpTargetErr	FFh	TxtpTargetErr[7:0]							
176	TxtpChkErrCnt	N/A	TxtpChkErrCnt[7:0]							
177	TxtpChkErrCnt	N/A	TxtpChkErrCnt[15:8]							
178	TxtpChkErrCnt	N/A	TxtpChkErrCnt[23:16]							
179	TxtpChkErrCnt	N/A	TxtpChkErrCnt[31:24]							
180	TxtpChkErrCnt	N/A	TxtpChkErrCnt[39:32]							
181	TxtpChkErr	N/A	TxtpChkErr[7:0]							
184	OutputLevel	F5h	HsPowerCtl[1:0]		Reserved	Reserved	Reserved	outLevel[2:0]		
205	TxPDGainAcq	8Dh	TxPDGainAcq[2:0]			Reserved	Reserved	Reserved	Reserved	Reserved
226	TxLoopFAcq	1Eh	TxLoop FAcqCtl	TxLoopFAcq[6:0]						

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Register 0. Part Identifier

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Identifier[7:0]							
Type	R							

Reset settings = 0100 0000

Bit	Name	Function
7:0	Identifier[7:0]	Second and least significant digit of the device part number (40).

Register 1. Part Identifier

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Revision[3:0]				Identifier[3:0]			
Type	R				R			

Reset settings = 0011 0000

Bit	Name	Function
7:4	Revision[3:0]	Die revision (Revision D = 3 decimal).
3:0	Identifier[3:0]	Third digit of the device number (0).

Register 2. ChipConfig1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	losOpenDrain	intOpenDrain	intEnable	spiOpenDrain	Reserved	lineside Loopback	XFI Loopback	refClkFreq
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0101 1000

Bit	Name	Function
7	losOpenDrain	**RX LOS Output Select.** 0 = CMOS output. 1 = Open Drain.
6	intOpenDrain	**Interrupt Pin Drive Select.** 0 = CMOS output. 1 = Open Drain.
5	intEnable	**Interrupt Enable.** 0 = Interrupts disabled. 1 = Interrupts enabled.
4	spiOpenDrain	**SD Pin Drive Configuration.** 0 = CMOS. 1 = Open Drain.
3	Reserved	Do not change. Must only write a 1 to this bit.
2	lineside Loopback	**Lineside Loopback Mode Control.** 0 = Normal operation. 1 = Optical data loopback.
1	XFILoopback	**XFI Loopback Mode Control.** 0 = Normal operation. 1 = XFI loopback.
0	refClkFreq	**Reference Clock Frequency Select.** 0 = 155 MHz. 1 = 622 MHz.

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Register 3. RxChipConfig2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name								RxPdn
Type	R	R	R	R	R	R	R	R/W

Reset settings = 0000 0000

Bit	Name	Function
7:1	Reserved	Read returns zero.
0	RxPdn	**Receiver Power Down.** 0 = Normal operation. 1 = Receiver powered down. A hard recal must be performed to calibrate all circuits (RX hardRecal at Register 8, Bit 3) when the receiver is returned to normal operation after a power down.

Register 4. RxintMask

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Interrupt.** 0 = Unmasked. Reference clock LOS generates an alarm on the Interrupt output pin (pin 16) if interrupts are enabled. (intEnable = 1) 1 = refLOS alarm is ignored.
5	LOS	**Loss of Signal Interrupt.** 0 = Unmasked. LOS generates an alarm on the Interrupt output (pin 16) if interrupts are enabled. (intEnable = 1) 1 = LOS alarm is ignored.
4	LOL	**Loss of Lock Interrupt.** 0 = Unmasked. LOL generates an alarm on the Interrupt output (pin 16) if interrupts are enabled. (intEnable = 1) 1 = LOL alarm is ignored.
3	fifoErr	**Receiver FIFO Error Interrupt.** 0 = Unmasked. fifoErr generates an alarm on the Interrupt output (pin 16) if interrupts are enabled (intEnable = 1). Will always read back zero if FIFOAutoReset is active (Reg28[1] = 1). 1 = fifoErr alarm is ignored.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm Interrupt.** 0 = Unmasked. tpErrAlarm generates an alarm on the Interrupt output (pin 16) if interrupts are enabled. (intEnable = 1) 1 = tpErrAlarm is ignored.
1	tpSyncLos	**Test Pattern Checker Loss of Sync Interrupt.** 0 = Unmasked. tpSyncLos generates an alarm on the Interrupt output (pin 16) if interrupts are enabled. (initEnable = 1) 1 = tpSyncLos alarm is ignored.
0	sqmAlarm	**Signal Quality Monitor Alarm Interrupt.** 0 = Unmasked. sqmAlarm generates an alarm on the Interrupt output (pin 16) if interrupts are enabled. (intEnable = 1) 1 = sqmAlarm is ignored.

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Register 5. RxintStatus (Sticky Bits)

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Interrupt.** A latched version of the refLOS alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
5	LOS	**Loss of Signal Interrupt.** A latched version of the LOS alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
4	LOL	**Loss of Lock Interrupt.** A latched version of the LOL alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
3	fifoErr	**Receiver FIFO Error Interrupt.** A latched version of the fifoErr alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm Interrupt.** A latched version of the tpErrAlarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
1	tpSyncLos	**Test Pattern Checker Loss of Sync Interrupt.** A latched version of the tpSyncLos alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
0	sqmAlarm	**Signal Quality Monitor Alarm Interrupt.** A latched version of the sqmAlarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.

Register 6. RxCmuConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	cmuBandwidth[3:0]				Reserved	Reserved		
Type	R/W				R	R/W		

Reset settings = 0100 0000

Bit	Name	Function
7:4	cmuBandwidth[3:0]	**RxCMU Jitter Transfer Bandwidth.** 0000 = Not supported 0001 = Not supported 0010 = Not supported 0011 = Not supported 0100 = 380 kHz Default 0101 = Not supported 0110 = Not supported
3	Reserved	Read returns zero.
2:0	Reserved	Do not change; must only write 000 to these bits.

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Register 7. RxConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		lolOnLOS	ltrOnLOS	CDRLTDATA	uselolMode	lolMode	ltr	rxRefclkEn
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0001 0101

Bit	Name	Function
7	Reserved	Read returns zero.
6	lolOnLOS	**Loss of Lock on Loss of Signal.** 0 = Normal LOL operation. 1 = Assert loss of lock on a loss of signal condition.
5	ltrOnLOS	**Lock to Reference on Loss of Signal.** 0 = Normal LTR operation. 1 = Assert lock to reference on a loss of signal condition.
4	CDRLTDATA	**CDR Lock Acquisition Options when Lock to Reference is enabled.** 0 = CDR phase locks to reference clock. 1 = CDR continues to attempt to lock to data.
3	uselolMode	**Loss of Lock Mode Overwrite.** 0 = Frequency LOL if RX ltr = 1 and SQM LOL if RX ltr = 0. The LOL mode can be read from the lolmode bit 2. 1 = LOL is based on lolMode bit2.
2	lolMode	**Loss of Lock Mode (This bit is only used if bit3 = 1).** 0 = LOL bases on frequency difference between the reference clock and received data. 1 = LOL bases on SQM.
1	ltr	**Lock to Reference.** 0 = Normal operation. 1 = RX CMU locked to reference clock.
0	rxRefclkEn	**Receiver Reference Clock Enable.** 0 = Reference clock not used by the receiver (incorrect rate). 1 = Reference clock applied to the receiver.

Register 8. RxCalConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name					hardRecal	VCOCAL[1:0]		swReset
Type	R	R	R	R	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7:4	Reserved	Read returns zero.
3	hardRecal	**Force Recalibarions.** 0 = Normal operation 1 = Initiate all calibrations of internal circuits and do not reset all RX registers. Bit is cleared upon completion of calibrations.
2:1	VCOCAL[1:0]	**Receive VCO Calibration Modes.** 00 (Default) = Automatic detection of the reference clock is enabled. If the reference clock is present and rxRefclkEn (Register 7, bit 0), it will be used to center the internal VCO pull range at the beginning of the lock acquisition process. Otherwise, the entire VCO frequency range will be swept. 01 = Enable referenceless operation. The entire VCO frequency range will be swept during the CDR lock acquisition process regardless of the presence of the reference clock. 10 = Enable reference operation. The internal VCO pull range will be centered with the reference clock frequency. 11 = Invalid mode. Note that receive LOL will always be on. **Note:** VCOCAL[1:0] must be set to reference (10b) or auto mode (00b) when part is configured to be in XFI loopback mode. **Note:** When Rx VCOCAL = x0b and a valid reference clock is present and rxRefclkEn = 1 (reg7[0]), registers 77 and 98 must not be written to. See section "5.8.1.1. Dynamic Register Control".
0	swReset	**Reset.** 0 = Normal operation. 1 = Reset all RX side registers. Bit is cleared upon completion of reset.

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Register 9. RxAlarmStatus

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R	R	R	R	R	R	R

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Alarm.** Loss of signal on the reference clock input, based on a coarse deviation in frequency.
5	LOS	**Loss of Signal Alarm.** Loss of signal on the receiver input. **Note:** This bit is the logical OR of the Analog LOS (aLOS), Digital LOS (dLOS) and Signal Quality Monitor LOS (sqmLOS) alarms, which can be enabled individually by programming Register 10.
4	LOL	**Loss of Lock Alarm.** The receiver PLL has lost lock with the input signal. Reflects the state of the RX_LOL pin (pin 2).
3	fifoErr	**Receiver FIFO Error Alarm.** The receiver FIFO has overflowed or underflowed. If FIFOAutoReset is active (Reg28[1] = 1), this bit is automatically cleared when a FIFO over/under flow has occurred.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm.** The receiver test pattern checker has reached the predetermined error count set in Register 47.
1	tpSyncLos	**Test Pattern Checker Loss-of-Sync Alarm.** The receiver test pattern checker has lost sync with the pattern. When the test pattern checker is disabled, this bit is automatically set to 0. When the test pattern checker is enabled, an "I" in this location means the test pattern checker has lost synchronization between the expected pattern and the received pattern.
0	sqmAlarm	**Signal Quality Monitor Alarm.** The internal signal quality monitor value has met the predetermined threshold value.

Register 10. RxLosCtrl

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name					sqmLosEn	dLosEn[1:0]		aLosEn
Type	R	R	R	R	R/W	R/W		R/W

Reset settings = 0000 0001

Bit	Name	Function
7:4	Reserved	Read returns zero.
3	sqmLosEn	**Signal Quality Monitor LOS Enable.** 0 = Disabled. 1 = Signal Quality Monitor alarm causes an LOS alarm.
2:1	dLosEn[1:0]	**Digital LOS Enable Mode.** 00 = Disabled. 01 = Digital LOS alarm is based on the consecutive number of zeros programmed in Register 17. 10 = Digital LOS alarm is based on the consecutive number of ones programmed in Register 17. 11 = Digital LOS alarm is based on the consecutive number of either zeros or ones programmed in Register 17.
0	aLosEn	**Analog LOS Enable.** 0 = Disabled. 1 = Analog LOS alarm active.

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Register 11. RxLosStatus

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				sqmLOS	dLOSlastTrigger	dLOS	aLOS	LOS
Type	R	R	R	R	R	R	R	R

Reset settings = 0000 0000

Bit	Name	Function
7:5	Reserved	Read returns zero.
4	sqmLOS	**Signal Quality Monitor LOS.** When the internal signal quality monitor (Reg25) is less than the threshold in registers 26 and 27, this bit will be high. If a LOL (Reg9[4]) is high, this bit is forced to a 0.
3	dLOSlastTrigger	**Digital Loss of Signal Last Trigger.** Holds a zero or one depending on which bit caused the last digital LOS event.
2	dLOS	**Digital Loss of Signal.** Digital LOS event status bit.
1	aLOS	**Analog Loss of Signal.** Analog LOS event status bit.
0	LOS	**Loss of Signal.** Reflects the state of the RX_LOS pin (pin 3).

Register 12. aLosThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	aLosThresh[7:0]							
Type	R/W							

Reset settings = 0000 1111

Bit	Name	Function
7:0	aLosThresh[7:0]	**Analog Loss of Signal Threshold.** Lower 8 bits of the analog LOS threshold value in mVppd. **Note:** Maximum analog LOS threshold is 400 mVppd.

Register 13. aLosThresh2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	aLosHyst[3:0]				Reserved		aLosThresh[9:8]	
Type	R/W				R		R/W	

Reset settings = 0111 0000

Bit	Name	Function
7:4	aLosHyst[3:0]	**Analog Loss of Signal Hysteresis.** Analog LOS deassert value = (aLosHyst + 16)/16 * aLosThresh in mVppd.
3:2	Reserved	Read returns zero.
1:0	aLosThresh[9:8]	**Analog Loss of Signal Threshold.** Upper two bits of the analog threshold value in mVppd. **Notes:** **1.** Maximum analog LOS threshold is 400 mVppd. **2.** Combined with Register 12, 00 0000 1010 = 10 mVppd minimum. 01 1001 0000 = 400 mVppd maximum.

Register 15. peakDet

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	peakDet[7:0]							
Type	R							

Reset settings = 0000 0000

Bit	Name	Function
7:0	peakDet[7:0]	**Peak Detector Signal Amplitude.** Least significant byte of the receiver peak detector signal amplitude in mV. This register should be read before register 16.

Register 16. peakDet

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name							peakDet[1:0]	
Type	R						R	

Reset settings = 0000 0000

Bit	Name	Function
7:2	Reserved	Read returns zero.
1:0	peakDet[9:8]	**Peak Detector Signal Amplitude.** Most significant two bits of the receiver peak detector signal amplitude in mV. This register should be read after reading Register 15. **Note:** Combined with Register 15, 00 0000 0000 = 0 mV 11 1111 1111 = 1022 mV

Register 17. RxdLosAssertThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxdLosAssertThresh[7:0]							
Type	R/W							

Reset settings = 0000 0101

Bit	Name	Function
7:0	RxdLosAssertThresh[7:0]	**Receiver Digital Loss of Signal Assert Threshold.** The number of consecutive identical digits before digital LOS is asserted. Assert threshold in bits = (RxdLosAssertThresh x 5 + 2) x 1024. See Figure 10 on page 20.

Register 18. RxdLosClearThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxdLosClearThresh[7:0]							
Type	R/W							

Reset settings = 0110 0000

Bit	Name	Function
7:0	RxdLosClearThresh[7:0]	**Receiver Digital Loss of Signal Clear Threshold.** Clear threshold that releases a digital LOS event. The number of consecutive 1024 bit fields with at least one transition that are required to clear dLos is calculated as RxdLos clearthresh x 16 + 1. See Figure 11 on page 21.

Register 20. sliceConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved					sliceEn[2:0]		
Type	R/W				R	R/W		

Reset settings = 0001 0001

Bit	Name	Function
7:4	Reserved	Do not change; must only write 0001 to these bits.
3	Reserved	Read returns zero.
2:0	sliceEn[2:0]	**Slice Mode Enable.** 000 = Slice disabled. 001 = Autoslice enabled. 010 = Constant duty cycle slice enabled. 011 = Proportional slice enabled. 100 = Absolute slice mode enabled.

Register 21. sliceLvl

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	sliceLvl[7:0]							
Type	R/W							

Reset settings = 0000 0000

Bit	Name	Function
7:0	sliceLvl[7:0]	**Slice Level.** Least significant byte of slice level setting. 2's compliment signed value. Absolute mode: 7FFFH = maximum slice offset = +240 mV 8000H = minimum slice offset = –240 mV Proportional Mode:Threshold = 50% + sliceLvl/65536 x 100 Constant Duty Cycle Mode:7A7AH = maximum duty cycle = ~74% 8586H = minimum duty cycle = ~26% $$\text{Duty Cycle} = 50\% + \frac{\text{sliceLvl[15:0]}}{1310}\%$$

Register 22. sliceLvl

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	sliceLvl[15:8]							
Type	R/W							

Reset settings = 0000 0000

Bit	Name	Function
15:8	sliceLvl[15:8]	**Slice Level.** Most significant byte of slice level setting. 2s compliment signed value. Absolute mode: 7FFFH = maximum slice offset = +240 mV 8000H = minimum slice offset = –240 mV Proportional Mode:Threshold = 50% + sliceLvl/65536 x 100 Constant Duty Cycle Mode:7A7AH = maximum duty cycle = ~74% 8586H = minimum duty cycle = ~26% $$\text{Duty Cycle} = 50\% + \frac{\text{sliceLvl[15:0]}}{1310}\%$$ **Note:** The slice level defined by Register 22 and 21 gets updated together after Register 21, the least significant byte of the 16-bit field, has been written. Prior to writing Register 21, the value written to Register 22 is stored into a mirroring register first. Any read to Register 22 prior to writing Register 21 will not return the intended value.

Register 24. RxphaseAdjust

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		RxphaseAdjust[6:0]						
Type	R	R/W						

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6:0	RxphaseAdjust[6:0]	**Receiver Phase Adjust.** Programmable range will cover at least –12 to 12 ps. The transfer function from Register 24 to the actual picoseconds of phase shift is highly variable. Value is signed, in 2s complement format: 100 0001 $\leq$ –12 ps 011 1111 $\geq$ 12 ps

Register 25. RxSqmValue

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			RxSqmValue[5:0]					
Type	R		R					

Reset settings = undefined

Bit	Name	Function
7:6	Reserved	Read returns zero.
5:0	RxSqmValue[5:0]	**Receiver Signal Quality Monitor Value.** Measured value of the magnitude of the received signal's horizontal eye opening. 00 0000 = minimum 11 1111 = maximum

Register 26. RxSqmConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				RxSqmThresh[5:0]				
Type				R/W			R/W	R/W

Reset settings = 0010 1001

Bit	Name	Function
7:2	RxSqmThresh[5:0]	**Receiver Signal Quality Monitor Threshold.** Threshold used to assert SQM LOS alarm. 00 0000 = 0 (decimal) 11 1111 = 63 (decimal) **Note:** Default = 10 (decimal)
1	Reserved	Do not change; must only write 0 to this bit.
0	Reserved	Do not change; must only write 1 to this bit.

Register 27. RxSqmDeassertThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			RxSqmDeassertThresh[5:0]					
Type	R		R/W					

Reset settings = 0000 1111

Bit	Name	Function
7:6	Reserved	Read returns zero.
5:0	RxSqmDeassertThresh[5:0]	**Receiver Signal Quality Monitor Deassert Threshold.** Value where the SQM alarm is removed. 00 0000 = 0 (decimal) 11 1111 = 63 (decimal) **Note:** Default = 15 (decimal)

Register 28. RxdPathConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		dinvert	clkOnLOS	SquelchO-nRxLOL	SquelchO-nRxLOS	Squelch	FIFOAuto Reset	FIFOReset
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0010

Bit	Name	Function
7	Reserved	Read returns zero.
6	dinvert	**Data Invert.** 0 = Normal operation. 1 = RD+ and RD– outputs (pins 19, 18) inverted.
5	clkOnLOS	**Clock Output on Receive Loss of Signal.** 0 = Normal operation. 1 = 622 MHz clock output on RD+ and RD– on receiver LOS condition.
4	SquelchOnRxLOL	**Data Squelch on Receive Loss of Lock.** 0 = Normal operation. 1 = Squelch RD+ and RD– outputs (pins 19, 18) on receiver Loss of Lock condition.
3	SquelchOnRxLOS	**Data Squelch on Receive Loss of Signal.** 0 = Normal operation. 1 = Squelch RD+ and RD– outputs (pins 19, 18) on receiver Loss of Signal condition.
2	Squelch	**Data Squelch.** 0 = Normal operation. 1 = Squelch RD+ and RD– outputs (pins 19, 18).
1	FIFOAutoReset	**FIFO Auto Reset.** 0 = No reset of receive FIFO on FIFO error. 1 = Automatically reset receive FIFO on FIFO underflow or overflow and clear fifoerr bit in RXintMask register (Reg4[1]). FIFO pointer is reset to center value and FIFO is cleared.
0	FIFOReset	**FIFO Reset.** 0 = Normal operation. 1 = Reset receive FIFO. FIFO pointer is reset to center value and FIFO is cleared.

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Register 29. RxtpSel

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	tpChkInvert	tpChkSel[2:0]			tpGenInvert	tpGenSel[2:0]		
Type	R/W	R/W			R/W	R/W		

Reset settings = 0000 0000

Bit	Name	Function
7	tpChkInvert	**Test Pattern Checker Data Invert.** 0 = Normal operation. 1 = Invert data applied to test pattern checker.
6:4	tpChkSel[2:0]	**Test Pattern Checker Mode Select.** 000 = Pattern checker disabled. 001 = Check for PRBS7 pattern. 010 = Check for PRBS31 pattern. 011 = Check for 64 bit user defined pattern.
3	tpGenInvert	**Test Pattern Generator Data Invert.** 0 = Normal operation. 1 = Invert generated pattern.
2:0	tpGenSel[2:0]	**Test Pattern Generator Mode Select.** 000 = Pattern generator disabled. 001 = Generate PRBS7 pattern. 010 = Generate PRBS31 pattern. 011 = Generate 64 bit user defined pattern. **Notes:** 1. Users *cannot* switch from generating PRBS31 to PRBS7 directly. tpGenSel should be set to 000 (pattern generator disabled) *before* switching to PRBS7. Note that this only applies to switching from PRBS31 to PRBS7. 2. The PRBS31pattern is inverted as per Section 5.8 of O.150.

Register 30. RxtpChkConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name						tpSyncMask	tpTimeBase[1:0]	
Type	R					R/W	R/W	

Reset settings = 0000 0010

Bit	Name	Function
7:3	Reserved	Read returns zero.
2	tpSyncMask	**Test Pattern Checker Sync Mask.** 0 = Normal operation. 1 = After synchronization has been achieved a loss of sync will not be reported and the pattern checker will not be reset.
1:0	tpTimeBase[1:0]	**Test Pattern Checker Timebase.** The time or number of bits over which the pattern will be checked. 00 = continuous (infinite). 01 = $(2^{20} - 1024)$ bits 10 = $(2^{30} - 1024)$ bits 11 = $(2^{40} - 1024)$ bits

Register 31. RxtpArbGenPtn

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxtpArbGenPtn[7:0]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[7:0]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

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Register 32. RxtpArbGenPtn

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	RxtpArbGenPtn[15:8]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[15:8]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 33. RxtpArbGenPtn

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name	RxtpArbGenPtn[23:16]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[23:16]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 34. RxtpArbGenPtn

Bit	D31	D30	D29	D28	D27	D26	D25	D24
Name	RxtpArbGenPtn[31:24]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[31:24]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 35. RxtpArbGenPtn

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name	RxtpArbGenPtn[39:32]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[39:32]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 36. RxtpArbGenPtn

Bit	D47	D46	D45	D44	D43	D42	D41	D40
Name	RxtpArbGenPtn[47:40]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[47:40]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 37. RxtpArbGenPtn

Bit	D55	D54	D53	D52	D51	D50	D49	D48
Name	RxtpArbGenPtn[55:48]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[55:48]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

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Register 38. RxtpArbGenPtn

Bit	D63	D62	D61	D60	D59	D58	D57	D56
Name	RxtpArbGenPtn[63:56]						
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbGenPtn[63:56]	**Receiver Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 31 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 38 is the MSB. The transmit sequence is from LSB to MSB.

Register 39. RxtpArbChkPtn

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxtpArbChkPtn[7:0]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[7:0]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 40. RxtpArbChkPtn

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	RxtpArbChkPtn[15:8]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[15:8]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 41. RxtpArbChkPtn

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name	RxtpArbChkPtn[23:16]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[23:16]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 42. RxtpArbChkPtn

Bit	D31	D30	D29	D28	D27	D26	D25	D24
Name	RxtpArbChkPtn[31:24]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[31:24]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 43. RxtpArbChkPtn

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name	RxtpArbChkPtn[39:32]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[39:32]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 44. RxtpArbChkPtn

Bit	D47	D46	D45	D44	D43	D42	D41	D40
Name				RxtpArbChkPtn[47:40]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[47:40]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 45. RxtpArbChkPtn

Bit	D55	D54	D53	D52	D51	D50	D49	D48
Name				RxtpArbChkPtn[55:48]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[55:48]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 46. RxtpArbChkPtn

Bit	D63	D62	D61	D60	D59	D58	D57	D56
Name				RxtpArbChkPtn[63:56]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	RxtpArbChkPtn[63:56]	**Receiver Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 39 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 46 is the MSB. The receive sequence is from LSB to MSB.

Register 47. RxtpTargetErr

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxtpTargetErr[7:0]							
Type	R/W							

Reset settings = 1111 1111

Bit	Name	Function
7:0	RxtpTargetErr[7:0]	**Receiver Test Pattern Checker Target Error Count.** If the value in the RxtpChkErrCnt register (register 53) exceeds this target error count, an interrupt will be generated. The value is represented as an 8-bit floating point number. Mantissa = bits[7:4] Exponent = bits[3:0] Error count = (Mantissa/16) x 16^{Exponent} 0000 0000 = 0 (decimal) 1111 1111 = (15/16) x 16^{15} (decimal) **Note:** This register value does not represent a target bit error rate (BER). Rather, it is a target bit error count for the period defined by tpTimeBase[1:0].

Register 48. RxtpChkErrCnt (LSB of a 40-bit Register)

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxtpChkErrCnt[7:0]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErrCnt[7:0]	**Receiver Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte (LSB) latches the upper bytes.

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Register 49. RxtpChkErrCnt (40-bit Register)

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	RxtpChkErrCnt[15:8]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErrCnt[15:8]	**Receiver Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 50. RxtpChkErrCnt (40-bit Register)

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name	RxtpChkErrCnt[23:16]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErrCnt[23:16]	**Receiver Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 51. RxtpChkErrCnt (40-bit Register)

Bit	D31	D30	D29	D28	D27	D62	D25	D24
Name	RxtpChkErrCnt[31:24]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErrCnt[31:24]	**Receiver Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 52. RxtpChkErrCnt (MSB of a 40-bit Register)

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name	RxtpChkErrCnt[39:32]
Type	R

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErrCnt[39:32]	**Receiver Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes. **Note:** Combined registers 48 to 52, 0000000000 = 0 (decimal) FFFFFFFFFF = $2^{40} - 1$ (decimal)

Register 53. RxtpChkErr

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxtpChkErr[7:0]
Type	R

Reset settings = undefined

Bit	Name	Function
7:0	RxtpChkErr[7:0]	**Receiver Test Pattern Checker Error.** Measured error count in 8-bit floating point notation. The contents of this register are an alternative format to the RxtpChkErrCnt. Mantissa = bits [7:4] Exponent = bits [3:0] Error count = (Mantissa/16) x $16^{Exponent}$ 0000 0000 = 0 (decimal) 1111 1111 = (15/16) x 16^{15} (decimal)

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Register 56. OutputLevel

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved		Reserved			outLevel[2:0]		
Type	R/W		R/W					

Reset settings = 1111 0101

Bit	Name	Function
7:6	Reserved	Do not change; must only write 11 to these bits.
5:3	Reserved	These bits are not user defined, and writes to these bits are ignored.
2:0	outLevel[2:0]	**RD output drive level.** 000 = 100 mVppd 001 = 200 mVppd 010 = 300 mVppd 011 = 400 mVppd 100 = 500 mVppd 101 = 600 mVppd 110 = 700 mVppd 111 = 800 mVppd

Register 77. RxPDGainAcq

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxPDGainAcq[2:0]			Reserved				
Type	R/W			R/W				

Reset Settings = 1000 1101

Bit	Name	Function
7:5	RxPDGainAcq[2:0]	RX phase detector gain during acquisition. Note that these bits require a one time write of 000b after a power up or a software reset.
4:0	Reserved	Reserved; must only write 01101b to these bits.

Register 84. RxEqConfig1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		RxEqGain				Reserved		
Type		R/W				R/W		

Reset settings = 1010 0001

Bit	Name	Function
7:5	RxEqGain	**Low-Frequency Gain in the Receiver Equalizer Frequency Response.** 000 0 dB (max gain) 001 −2 dB 010 −2 dB (same as 001 setting) 011 −3 dB 100 −4 dB 101 −5 dB 110 −6 dB 111 −7 dB (min gain)
4:0	Reserved	Do not change; must only write 00001 to these bits.

Register 85. RxEqConfig2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		RxEqHFBoost				Reserved		
Type		R/W				R/W		

Reset settings = 1110 0000

Bit	Name	Function
7:5	RxEqHFBoost	**High-Frequency Boost in the Receiver Equalizer Frequency Response.** 000 0 dB (min boost setting) 001 1 dB 010 2 dB 011 3 dB 100 4 dB 101 5 dB 110 6 dB 111 7 dB (max boost setting)
4:0	Reserved	Do not change; must only write 00000 to these bits.

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Register 98. RxLoopFAcq

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	RxLoopFAcqCtl	|	|	|	RxLoopFAcq[6:0]	|	|	|
Type	R/W	|	|	|	R/W	|	|	|

Reset settings = 0001 1110

Bit	Name	Function
7	RxLoopFAcqCtl	**RX Acquistion Loop Filter Override.** 1 = Use value written in Bit [6:0]. Set to 1 only when RX LOL is asserted. 0 = Use internally generated value. Set to 0 when RX LOL is deasserted.
6:0	RxLoopFAcq[6:0]	**RX Loop Filter Setting for Acquisition.** RX Loop filter override setting to be used during acquisiton. Set to 001 1000b when RX LOL is asserted and to 000 0000b when RX LOL is deasserted. Note that any read back may not return the last written value.

Register 106. sqmLOLThreshWrt

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	sqmLOLThreshWrt	Reserved	|	|	sqmLOLThreshAdd[5:0]	|	|	|
Type	R/W	R/W	|	|	R/W	|	|	|

Reset settings = 0000 0000

Bit	Name	Function
7	sqmLOLThreshWrt	Self-clearning strobe bit to apply sqmLOL Threshold from registers 107,108,109. This is an indexing address register and requires the sqmLOL threshold address to be set before writing data in the appropriate indexed data registers. For this reason, it must first be written to 04h, followed by 84h. Refer to Section 5.8.1 for more information about this register.
6	Reserved	Do not change; must only write 0 to this bit.
5:0	sqmLOL ThreshAdd[5:0]	Address for sqmLOL Threshold. Must be set to 4.

Register 107. sqmLOLThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	sqmLOLThresh[0]	Reserved						
Type	R/W	R/W						

Reset settings = 0000 0000

Bit	Name	Function
7	sqmLOLThresh[0]	**SQM LOL Threshold.**
		Least significant bit of 14 bit SQM LOL Threshold setting; value is unsigned integer value. RxLOL is asserted when jitter measure exceeds value in sqmLOLThresh[13:0] and is deasserted when jitter measure is below the threshold. Refer to "5.8.1. SQM LOL" on page 24 for more information about this register.
6:0	Reserved	Reserved. Should be written to 010 0000b.

Register 108. sqmLOLThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	sqmLOLThresh[8:1]							
Type	R/W	R/W						

Reset settings = 0000 0000

Bit	Name	Function
7:0	sqmLOLThresh[8:1]	**SQM LOL Threshold.**
		Bit [8:1] of sqmLOL Threshold setting; value is unsigned integer value. RxLOL is asserted when jitter measure exceeds value in sqmLOLThresh[13:0] and is deasserted when jitter measure is below the threshold. Refer to Section 5.8.1 for more information about this register.

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Register 109. sqmLOLThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved			sqmLOLThresh[13:9]				
Type	R/W			R/W				

Reset settings = 0000 0000

Bit	Name	Function
7:5	Reserved	Reserved. Should be written to 101b.
4:0	sqmLOLThresh[13:9]	**SQM LOL Threshold.** Bit [13:9] of sqmLOL Threshold setting; value is unsigned integer value. RxLOL is asserted when jitter measure exceeds value in sqmLOLThresh[13:0] and is deasserted when jitter measure is below the threshold. Refer to Section 5.8.1 for more information about this register.

Register 131. TxChipConfig2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name							Reserved	TxPdn
Type	R			R/W			R	R/W

Reset settings = 0010 0010

Bit	Name	Function
7	Reserved	Read returns zero.
6:1	Reserved	Do not change; must only write 010001 to these bits.
0	TxPdn	**Transmitter Power Down.** 0 = Normal operation. 1 = Transmitter powered down. A hard recal must be performed to calibrate all circuits (TX hardRecal at Register 136, Bit 3) when the transmitter is returned to normal operation after a power down.

Register 132. TxintMask

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Interrupt.** 0 = Unmasked. Reference clock LOS generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). 1 = refLOS alarm is ignored.
5	LOS	**Loss of Signal Interrupt.** 0 = Unmasked. LOS generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). 1 = LOS alarm is ignored.
4	LOL	**Loss of Lock Interrupt.** 0 = Unmasked. LOL generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). 1 = LOL alarm is ignored.
3	fifoErr	**Transmitter FIFO Error Interrupt.** 0 = Unmasked. fifoErr generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). Will always read back zero if FIFOAutoReset is active (Reg156[1] = 1). 1 = fifoErr alarm is ignored.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm Interrupt.** 0 = Unmasked. tpErrAlarm generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). 1 = tpErrAlarm is ignored.
1	tpSyncLos	**Test Pattern Checker Loss of Sync Interrupt.** 0 = Unmasked. tpSyncLos generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (initEnable = 1). 1 = tpSyncLos alarm is ignored.
0	sqmAlarm	**Signal Quality Monitor Alarm Interrupt.** 0 = Unmasked. sqmAlarm generates an alarm on the Interrupt output pin (pin 12) if interrupts are enabled. (intEnable = 1). 1 = sqmAlarm is ignored.

Register 133. TxintStatus (Sticky Bits)

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Interrupt.** A latched version of the refLOS alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
5	LOS	**Loss of Signal Interrupt.** A latched version of the LOS alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
4	LOL	**Loss of Lock Interrupt.** A latched version of the LOL alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
3	fifoErr	**Transmitter FIFO Error Interrupt.** A latched version of the fifoErr alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm Interrupt.** A latched version of the tpErrAlarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
1	tpSyncLos	**Test Pattern Checker Loss of Sync Interrupt.** A latched version of the tpSyncLos alarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.
0	sqmAlarm	**Signal Quality Monitor Alarm Interrupt.** A latched version of the sqmAlarm status bit. An interrupt is generated if interrupts are enabled (intEnable = 1) and if not masked by the corresponding interrupt mask bit. The interrupt may be cleared by writing a zero to this bit position or by disabling interrupts.

Register 134. TxCmuConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	cmuBandwidth[3:0]					cmuMode[2:0]		
Type	R/W				R	R/W		

Reset settings = 0100 0000

Bit	Name	Function
7:4	cmuBandwidth[3:0]	**TxCMU Jitter Transfer Bandwidth.** 0000 = 180 Hz Valid for CMU modes 2 and 6. 0001 = 1.37 kHz Valid for CMU modes 2 and 6. 0010 Not supported. 0011 Not supported. 0100 = 380 kHz Valid for CMU modes 0 and 1 only. 0101 Not supported. 0110 Not supported. **Note:** A manual recalibration (hardRecal = 1 in Register 8) is required after a change in the CMU jitter transfer bandwidth.
3	Reserved	Read returns zero.
2:0	cmuMode[2:0]	**CMU Mode of Operation.** 000 = Referenceless Mode. 001 = Synchronous Reference Clock Mode (This is logically the same as setting the ltr bit (register135[1])). 010 = Asynchronous Reference Clock Mode. 110 = Looptime Mode with Clean Reference Clock. (Recommended CMU bandwidth = 1.37 kHz.) **Note:** An automatic recalibration is executed after a change in timing mode.

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Register 135. TxConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				CDRLTDATA	uselolMode	lolMode	ltr	
Type	R/W							R

Reset settings = 1001 0100

Bit	Name	Function
7:5	Reserved	Do not change; must only write 100 to these bits.
4	CDRLTDATA	**CDR Lock Acquisition Options when Lock to Reference is Enabled.** 0 = CDR phase locks to reference clock. 1 = CDR continues to attempt to lock to data.
3	uselolMode	Loss of Lock Mode Overwrite. 0 = Auto select LOL mode is based upon the tx CMU mode. The selected LOL mode can be read from the LOL mode bit 2. 1 = LOL is based on lolMode bit2.
2	lolMode	**Loss of Lock Mode (This bit is only used if bit3 = 1).** 0 = LOL bases on frequency difference between the reference clock and received data. 1 = LOL bases on SQM.
1	ltr	**Lock to Reference.** 0 = Normal operation. 1 = CMU is locked to reference clock.
0	Reserved	Read returns zero.

Register 136. TxCalConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name					hardRecal	VCOCAL[1:0]		swReset
Type	R	R	R	R	R/W	R/W	R/W	R/W

Reset settings = 0000 0000

Bit	Name	Function
7:4	Reserved	Read returns zero.
3	hardRecal	**Force Recalibrations.** 0 = Normal operation 1 = Initiate all calibrations of internal circuits and do not reset all TX registers. Bit is cleared upon completion of calibrations.
2:1	VCOCAL[1:0]	**Transmit VCO Calibration Modes.** 00 = (Default) Automatic detection of the reference clock is enabled. If the reference clock is present, it will be used to center the internal VCO pull range at the beginning of the lock acquisition process. Otherwise, the entire VCO frequency range will be swept. 01 = Enable referenceless operation. The entire VCO frequency range will be swept during the CDR lock acquisition process regardless of the presence of the reference clock. 10 = Enable reference operation. The internal VCO pull range will be centered with the reference clock frequency. 11 = Invalid mode. Note that transmit LOL will always be on. **Note:** VCOCAL[1:0] must be set to reference(10b) or auto mode (00b) when part is configured to be in Lineside loopback mode. **Note:** When Tx VCOCAL = x0b and a valid reference clock is present, registers 205 and 226 must not be written to. See section "6.4.1.1. Dynamic Register Control".
0	swReset	**Software Reset.** 0 = Normal operation. 1 = Reset all TX side registers. Bit is cleared upon completion of reset.

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Register 137. TxAlarmStatus

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		refLOS	LOS	LOL	fifoErr	tpErrAlarm	tpSyncLos	sqmAlarm
Type	R	R	R	R	R	R	R	R

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6	refLOS	**Reference Clock LOS Alarm.** Loss of signal on the reference clock input, based on a coarse deviation in frequency.
5	LOS	**Loss of Signal Alarm.** Loss of signal on the transmitter input. (TD)
4	LOL	**Loss of Lock Alarm.** The transmitter PLL has lost lock with the transmitter input signal. (TD)
3	fifoErr	**Transmitter FIFO Error Alarm.** The transmitter FIFO has overflowed or underflowed. If FIFOAutoReset is active (Reg156[1] = 1), this bit is automatically cleared when a FIFO over/under flow has occurred.
2	tpErrAlarm	**Test Pattern Generator/Checker Alarm.** The transmitter test pattern checker has reached the predetermined error count set in Register 175.
1	tpSyncLos	**Test Pattern Checker Loss-of-Sync Alarm.** The receiver test pattern checker has lost sync with the pattern. When the test pattern checker is disabled, this bit is automatically set to 0. When the test pattern checker is enabled, an "I" in this location means the test pattern checker has lost synchronization between the expected pattern and the received pattern.
0	sqmAlarm	**Signal Quality Monitor Alarm.** The internal signal quality monitor value has met the predetermined threshold value.

Register 138. TxLosCtrl

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name					sqmLosEn	dLosEn[1:0]		
Type	R				R/W	R/W		R

Reset settings = 0000 1110

Bit	Name	Function
7:4	Reserved	Read returns zero.
3	sqmLosEn	**Signal Quality Monitor LOS Enable.** 0 = Disabled. 1 = Signal Quality Monitor alarm causes an LOS alarm.
2:1	dLosEn[1:0]	**Digital LOS Enable Mode.** 00 = Disabled. 01 = Digital LOS alarm is based on the consecutive number of zeros programmed in Register 145. 10 = Digital LOS alarm is based on the consecutive number of either zeros or ones programmed in Register 145. 11 = Digital LOS alarm is based on the consecutive number of either zeros or ones programmed in Register 145.
0	Reserved	Read returns zero.

Register 139. TxLosStatus

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				sqmLOS	dLOSlastTrigger	dLOS		LOS
Type	R	R	R	R	R	R	R	R

Reset settings = 0000 0000

Bit	Name	Function
7:5	Reserved	Read returns zero.
4	sqmLOS	**Signal Quality Monitor LOS.** When the internal signal quality monitor (Reg153) is less than the threshold in registers 154 and 155, this bit will be high. If a LOL (Reg137[4]) is high, this bit is forced to a 0.
3	dLOSlastTrigger	**Digital Loss of Signal Last Trigger.** Holds a zero or one depending on which bit caused the last digital LOS event.
2	dLOS	**Digital Loss of Signal.** Digital LOS event status bit.
1	Reserved	Read returns zero.
0	LOS	**Loss of Signal.** LOS status bit.

Register 145. TxdLosAssertThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxdLosAssertThresh[7:0]							
Type	R/W							

Reset settings = 0000 0000

Bit	Name	Function
7:0	TxdLosAssertThresh	**Transmitter Digital Loss of Signal Assert Threshold.** The number of consecutive identical digits before digital LOS is asserted. Assert threshold in bits = (TxdLosAssertThresh x 5 + 2) x 1024. See Figure 10 on page 20.

Register 146. TxdLosClearThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxdLosClearThresh[7:0]							
Type	R/W							

Reset settings = 0110 0000

Bit	Name	Function
7:0	TxdLosClearThresh[7:0]	**Transmitter Digital Loss of Signal Clear Threshold.** Clear threshold that releases a digital LOS event. The number of consecutive 1024 bit fields with at least one transition that are required to clear dLos is calculated as RxdLos clearthresh x 16 + 1. See Figure 11 on page 21.

Register 152. TxphaseAdjust

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		TxphaseAdjust[6:0]						
Type	R	R/W						

Reset settings = 0000 0000

Bit	Name	Function
7	Reserved	Read returns zero.
6:0	TxphaseAdjust[6:0]	**Transmitter Phase Adjust.** Programmable range will cover at least –12 to 12 ps. The transfer function from Register # to actual ps of phase shift is highly variable. Value is signed, in 2s complement format: 100 0001 $\leq$ –12 ps 011 1111 $\geq$ 12 ps

Register 153. TxSqmValue

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			TxSqmValue[5:0]					
Type	R		R/W					

Reset settings = 0000 0000

Bit	Name	Function
7:6	Reserved	Read returns zero.
5:0	TxSqmValue[5:0]	**Transmitter Signal Quality Monitor Value.** Measured value of the magnitude of the transmitter's received signal horizontal eye opening. 00 0000 = minimum 11 1111 = maximum

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Register 154. TxSqmConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxSqmThresh[5:0]							
Type	R/W						R/W	R/W

Reset settings = 0000 0101

Bit	Name	Function
7:2	TxSqmThresh[5:0]	**Transmitter Signal Quality Monitor Threshold.** Threshold used to assert SQM LOS alarm. 00 0000 = 0 (decimal) 11 1111 = 63 (decimal) **Note:** Default = 1 (decimal)
1	Reserved	Do not change; must only write a 0 to this bit.
0	Reserved	Do not change; must only write a 1 to this bit.

Register 155. TxSqmDeassertThresh

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			TxSqmDeassertThresh[5:0]					
Type	R/W							

Reset settings = 0000 1000

Bit	Name	Function
7:6	Reserved	Read returns zero.
5:0	TxSqmDeassertThresh[5:0]	**Transmitter Signal Quality Monitor Deassert Threshold.** Value where the SQM alarm is removed. 00 0000 = 0 (decimal) 11 1111 = 63 (decimal) **Note:** Default = 8 (decimal)

Register 156. TxdPathConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		dinvert	clkOnLOS	SquelchOnTxLOL	SquelchOnTxLOS	Squelch	FIFOAutoReset	FIFOReset
Type	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset settings = 0000 0010

Bit	Name	Function
7	Reserved	Read returns zero.
6	dinvert	**Data Invert.** 0 = Normal operation. 1 = TXDOUT+ and TXDOUT− outputs (pins 30, 29) are inverted.
5	clkOnLOS	**Clock Output on Transmitter Loss of Signal.** 0 = Normal operation. 1 = 622 MHz clock output on TXDOUT+ and TXDOUT− on transmitter LOS condition.
4	SquelchOnTxLOL	**Data Squelch on Transmit Loss of Lock.** 0 = Normal operation. 1 = squelch TXDOUT output (pins 30, 29) on transmitter Loss of Lock condition.
3	SquelchOnTxLOS	**Data Squelch on Transmit Loss of Signal.** 0 = Normal operation. 1 = squelch TXDOUT output (pins 30, 29) on transmitter Loss of Signal condition.
2	Squelch	**Data Squelch.** 0 = Normal operation. 1 = squelch TXDOUT output (pins 30, 29).
1	FIFOAutoReset	**FIFO Auto Reset.** 0 = No reset of transmit FIFO on FIFO error. 1 = automatically reset transmit FIFO on FIFO underflow or overflow. FIFO pointer is reset to center value and FIFO is cleared.
0	FIFOReset	**FIFO Reset.** 0 = Normal operation. 1 = reset transmit FIFO. FIFO pointer is reset to center value and FIFO is cleared.

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Register 157. TxtpSel

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	tpChkInvert	tpChkSel[2:0]			tpGenInvert	tpGenSel[2:0]		
Type	R/W	R/W			R/W	R/W		

Reset settings = 0000 0000

Bit	Name	Function
7	tpChkInvert	**Test Pattern Checker Data Invert.** 0 = normal operation. 1 = invert data applied to test pattern checker.
6:4	tpChkSel[2:0]	**Test Pattern Checker Mode Select.** 000 = pattern checker disabled. 001 = check for PRBS7 pattern. 010 = check for PRBS31 pattern. 011 = check for 64 bit user defined pattern.
3	tpGenInvert	**Test Pattern Generator Data Invert.** 0 = normal operation. 1 = invert generated pattern.
2:0	tpGenSel[2:0]	**Test Pattern Generator Mode Select.** 000 = pattern generator disabled. 001 = generate PRBS7 pattern. 010 = generate PRBS31 pattern. 011 = generate 64 bit user defined pattern. **Notes:** **1.** Users *cannot* switch from generating PRBS31 to PRBS7 directly. tpGenSel should be set to 000 (pattern generator disabled) *before* switching to PRBS7. Note that this only applies to switching from PRBS31 to PRBS7. **2.** The PRBS31pattern is inverted as per Section 5.8 of O.150.

Register 158. TxtpChkConfig

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name						tpSyncMask	tpTimeBase[1:0]	
Type	R					R/W	R/W	

Reset settings = 0000 0010

Bit	Name	Function
7:3	Reserved	Read returns zero.
2	tpSyncMask	**Test Pattern Checker Sync Mask.** 0 = Report loss of sync, followed by reset. 1 = After synchronization has been achieved a loss of sync will not be reported and the pattern checker will not be reset.
1:0	tpTimeBase[1:0]	**Test Pattern Checker Timebase.** The time or number of bits over which the pattern will be checked. 00 = continuous (infinite). 01 = $(2^{20} – 1024)$ bits 10 = $(2^{30} – 1024)$ bits 11 = $(2^{40} – 1024)$ bits

Register 159. TxtpArbGenPtn

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxtpArbGenPtn[7:0]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[7:0]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 160. TxtpArbGenPtn

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	TxtpArbGenPtn[15:8]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[15:8]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 161. TxtpArbGenPtn

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name	TxtpArbGenPtn[23:16]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[23:16]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 162. TxtpArbGenPtn

Bit	D31	D30	D29	D28	D27	D26	D25	D24
Name	TxtpArbGenPtn[31:24]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[31:24]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 163. TxtpArbGenPtn

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name	TxtpArbGenPtn[39:32]
Type	R/W

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[39:32]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 164. TxtpArbGenPtn

Bit	D47	D46	D45	D44	D43	D42	D41	D40
Name	TxtpArbGenPtn[47:40]
Type	R/W

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[47:40]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 165. TxtpArbGenPtn

Bit	D55	D54	D53	D52	D51	D50	D49	D48
Name	TxtpArbGenPtn[55:48]
Type	R/W

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[55:48]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

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Register 166. TxtpArbGenPtn

Bit	D63	D62	D61	D60	D59	D58	D57	D56
Name	TxtpArbGenPtn[63:56]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbGenPtn[63:56]	**Transmitter Test Pattern Generator User Defined Pattern.** **Note:** Bit 0 in Register 159 is the LSB of the 64-bit user-defined pattern, and Bit 7 in Register 166 is the MSB. The transmit sequence is from LSB to MSB.

Register 167. TxtpArbChkPtn

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxtpArbChkPtn[7:0]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[7:0]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 168. TxtpArbChkPtn

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	TxtpArbChkPtn[15:8]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[15:8]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 169. TxtpArbChkPtn

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name				TxtpArbChkPtn[23:16]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[23:16]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 170. TxtpArbChkPtn

Bit	D31	D30	D29	D28	D27	D26	D25	D24
Name				TxtpArbChkPtn[31:24]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[31:24]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 171. TxtpArbChkPtn

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name				TxtpArbChkPtn[39:32]				
Type				R/W				

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[39:32]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

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Register 172. TxtpArbChkPtn

Bit	D47	D46	D45	D44	D43	D42	D41	D40
Name	TxtpArbChkPtn[47:40]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[47:40]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 173. TxtpArbChkPtn

Bit	D55	D54	D53	D52	D51	D50	D49	D48
Name	TxtpArbChkPtn[55:48]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[55:48]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 174. TxtpArbChkPtn

Bit	D63	D62	D61	D60	D59	D58	D57	D56
Name	TxtpArbChkPtn[63:56]							
Type	R/W							

Reset settings = 1010 1010

Bit	Name	Function
7:0	TxtpArbChkPtn[63:56]	**Transmitter Test Pattern Checker User Defined Pattern.** **Note:** Bit 0 in Register 167 is the LSB of the received 64-bit user-defined pattern, and Bit 7 in Register 174 is the MSB. The receive sequence is from LSB to MSB.

Register 175. TxtpTargetErr

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxtpTargetErr[7:0]							
Type	R/W							

Reset settings = 1111 1111

Bit	Name	Function
7:0	TxtpTargetErr[7:0]	**Transmitter Test Pattern Checker Target Error Count.** If the value in the TxtpChkErrCnt register (register 181) exceeds this target error count, an interrupt will be generated. The value is represented as an 8-bit floating point number. Mantissa = bits[7:4] Exponent = bits[3:0] Base = 16 0000 0000 = 0 (decimal) 0101 0111 = (5/16) x 16^7 (decimal) 1111 1111 = (15/16) x 16^{15}(decimal) This register value does not represent a target bit error rate (BER). Rather, it is a target bit error count for the period defined by tpTimeBase[1:0].

Register 176. TxtpChkErrCnt (LSB of a 40-bit Register)

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxtpChkErrCnt[7:0]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErrCnt[7:0]	**Transmitter Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte LSB latches the upper bytes.

Register 177. TxtpChkErrCnt (40-bit Register)

Bit	D15	D14	D13	D12	D11	D10	D9	D8
Name	TxtpChkErrCnt[15:8]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErrCnt[15:8]	**Transmitter Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 178. TxtpChkErrCnt (40-bit Register)

Bit	D23	D22	D21	D20	D19	D18	D17	D16
Name	TxtpChkErrCnt[23:16]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErrCnt[23:16]	**Transmitter Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 179. TxtpChkErrCnt (40-bit Register)

Bit	D31	D30	D29	D28	D27	D26	D25	D24
Name	TxtpChkErrCnt[31:24]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErrCnt[31:24]	**Transmitter Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes.

Register 180. TxtpChkErrCnt (MSB of a 40-bit Register)

Bit	D39	D38	D37	D36	D35	D34	D33	D32
Name	TxtpChkErrCnt[39:32]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErrCnt[39:32]	**Transmitter Test Pattern Checker Error Count.** When using a defined timebase, this register holds the error count from the last completed timebase. In the continuous timebase setting, the register holds the current running error count. Reading the least significant byte latches the upper bytes. **Note:** Combined registers 176 to 180, 0000000000 = 0 (decimal) FFFFFFFFFF = $2^{40} - 1$ (decimal)

Register 181. TxtpChkErr

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxtpChkErr[7:0]							
Type	R							

Reset settings = undefined

Bit	Name	Function
7:0	TxtpChkErr[7:0]	**Transmitter Test Pattern Checker Error.** Measured error count in 8-bit floating point notation. The content of this register is an alternative format to the TxtpChkErrCnt. Mantissa = bits [7:4] Exponent = bits [3:0] Error count = (Mantissa/16) x $16^{Exponent}$ 0000 0000 = 0 (decimal) 0101 0111 = (5/16) x 16^7 (decimal) 1111 1111 = (15/16) x 16^{15} (decimal)

Register 184. OutputLevel

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	HsPowerCtl[1:0]		Reserved			outLevel[2:0]		
Type	R/W		R/W			R/W		

Reset settings = 1111 0101

Bit	Name	Function
7:6	Reserved	Do not change; must only write 11 to these bits.
5:3	Reserved	These bits are not user defined, and writes to these bits are ignored.
2:0	outLevel[2:0]	**Output Level.** TXDOUT output drive level. 000 = 100 mVppd. 001 = 200 mVppd. 010 = 300 mVppd. 011 = 400 mVppd. 100 = 500 mVppd. 101 = 600 mVppd. 110 = 700 mVppd. 111 = 800 mVppd.

Register 205. TxPDGainAcq

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxPDGainAcq[2:0]			Reserved				
Type	R/W			R/W				

Reset settings = 1000 1101

Bit	Name	Function
7:5	TxPDGainAcq[2:0]	TX phase detector gain during acquisition. Note that these bits require a one time write of 000b after a power-up or a software reset.
4:0	Reserved	Reserved; must only write 01101b to these bits.

Register 226. TxLoopFAcq

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	TxLoopFAcqCtl	TxLoopFAcq[6:0]						
Type	R/W	R/W						

Reset settings = 0001 1110

Bit	Name	Function
7	TxLoopFAcqCtl	**TX Acquistion Loop Filter Override.** 1 = Use value written in Bit [6:0]. Set to 1 only when TX LOL is asserted. 0 = Use internally generated value. Set to 0 when TX LOL is deasserted.
6:0	TxLoopFAcq[6:0]	**TX Loop Filter Setting for Acquisition.** TX Loop filter override setting to be used during acquisiton. Set to 001 1000b when TX LOL is asserted and to 000 0000 when TX LOL is deasserted. Note that any read back may not return the last written value.

14. Pin Descriptions: Si5040
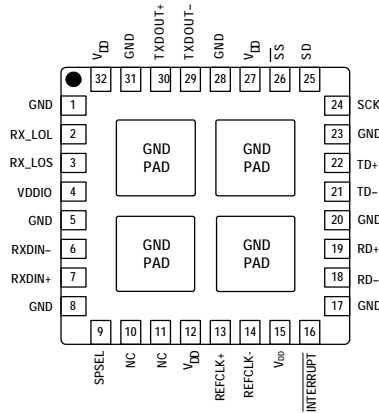
Figure 24. Si5040 Pin Configuration (Transparent Top View)

Table 13. Si5040 Pin Descriptions

Pin	Name	Type*	Level	Description
16	INTERRUPT	DO	LVTTL or Open Drain	**Interrupt (Active Low).** The interrupt output pin is provided to indicate potential fault conditions or changes in status. Interrupt sources are maskable by setting the Interrupt Mask register, and interrupt status is available from the Interrupt Status register. The interrupt function can be disabled in the Interrupt Enable bit. The interrupt pin can be configured via the Interrupt Output register as either an open drain output (default) or LVTTL output.
19,18	RD+ RD–	AO	Differential CML	**Receiver Data Output.** High-speed XFI-compliant receiver data output recovered from the RXDIN input.
13,14	REFCLK+ REFCLK–	AI	PECL	**Reference Clock Input.** A reference clock at this input is applied to the transmit CMU and to the receiver and transmitter CDRs. The use of a reference clock is optional. If the jitter performance of the external reference clock is acceptable, the Si5040 can be operated in CMU mode. In this mode, the CMU derives the line-rate clock by multiplying the clock frequency applied to the REFCLK inputs. If the REFCLK input is synchronous, the CMU multiplies the frequency by 64. The resulting line-rate is frequency-locked to the serial data. A FIFO in the data path accommodates any jitter differences between the serial data and the CMU line-rate clock.
7, 6	RXDIN+ RXDIN–	AI	Differential CML	**Receiver Data Input.** Data signal RD is recovered from the high-speed differential signal present on these pins. Data over the 9.8304 Gbps to 11.3 Gbps range is recovered.
*Note: TYPE: P = Power; AI = Analog Input; AO = Analog Output; DI = Digital Input; DO = Digital Output; DIO = Digital Input/Output.				

Table 13. Si5040 Pin Descriptions (Continued)

Pin	Name	Type*	Level	Description
2	RX_LOL	DO	LVTTL	**Receiver Loss of Lock (Active High).** This output is asserted when the receiver path is in the loss-of-lock state. If enabled in the receiver Interrupt Mask register, this event may cause an interrupt. This pin is reflected as bit 4 in Register 9. The latched version of this pin is in Register 5, bit 4. In the absence of an external reference, the lock detect circuitry uses a data quality measure to determine when frequency lock has been lost with the incoming data stream. This pin may also be programmed as a 622 MHz clock output that is synchronous to the data applied at the transmitter data (TD) input.
3	RX_LOS	DO	LVTTL	**Receiver Loss of Signal (Active High).** The RX_LOS output is asserted when a loss-of-signal condition occurs for Analog LOS, Digital Count LOS, or Signal Quality Monitor LOS.
22,21	TD+ TD–	AI	Differential CML	**Transmitter Data Input.** High-speed XFI-compliant transmitter data input.
30,29	TXDOUT+ TXDOUT–	AO	Differential CML	**Transmitter Data Output.** Data present at the TD input is retimed and output on the TXDOUT pins. Transmit range of operation is 9.8304 Gbps to 11.3 Gbps.
				Serial Port
24	SCK	DI	LVTTL	**Serial Clock.** Clock input for SPI-like and I^2C interface.
25	SD	DIO	LVTTL or Open Drain	**Serial Data.** Serial bidirectional data interface pin for the SPI-like or I^2C serial interface. This pin may be programmed as LVTTL or open drain (by default) in Register 2. When in LVTTL mode, the pin is initially in a high-impedance state.
9	SPSEL	DI	LVTTL	**SPI-like or I^2C Bus Select.** In a logic low state, this input selects I^2C mode of operation. In a logic high state, SPI-like mode of operation is selected. This input has a weak internal pullup.
26	$\overline{SS}$	DI	LVTTL	**Chip Select.** Chip select pin for SPI-like interface, active low. This input has a weak internal pullup. Note that this pin defines the I^2C when the serial interface is in I^2C mode. See "10.1. I2C Interface" on page 35 for details.

*Note: TYPE: P = Power; AI = Analog Input; AO = Analog Output; DI = Digital Input; DO = Digital Output; DIO = Digital Input/Output.

Table 13. Si5040 Pin Descriptions (Continued)

Pin	Name	Type*	Level	Description
Power and Ground				
1, 5, 8, 17, 20, 23, 28, 31	GND	P	GND	**Supply Ground.** Connect to system GND. Ensure a very low impedance path for optimal performance.
Paddle	GND	P	GND	**Paddle Ground.** Must connect to system GND. Ensure a very low impedance path for optimal performance.
12, 15, 27,32	VDD	P	1.8 VDC	**Supply Voltage.** Nominally 1.8 V. Upon power up,if VDD and VDDIO are not tied to the same supply voltage, the VDD voltage must achieve 1.5 V before VDDIO is applied and reaches .25 V. Failure to do so may cause the SD line to glitch low.
4	VDDIO	P	3.3 VDC or 1.8 VDC	**LVTTL I/O Supply Voltage.** Connect to either 1.8 or 3.3 V. When connected to 3.3 V, LVTTL compatible voltage swings are supported on the LVTTL inputs and LVTTL outputs of the device.
NC Pins				
11 10				**NC.** Do not connect to these pins.
*Note: TYPE: P = Power; AI = Analog Input; AO = Analog Output; DI = Digital Input; DO = Digital Output; DIO = Digital Input/Output.				

15. Ordering Guide

Part Number*	Package	Lead-Free	Temperature
Si5040-D-GM	32-lead LGA	Yes	–40 to 95 °C
*Note: Add an "R" at the end of the device number to denote the tape and reel option; 2500 quantity per reel.			

16. Package Outline: Si5040

Figure 25 illustrates the package details for the Si5040. Table 14 lists the values for the dimensions shown in the illustration.
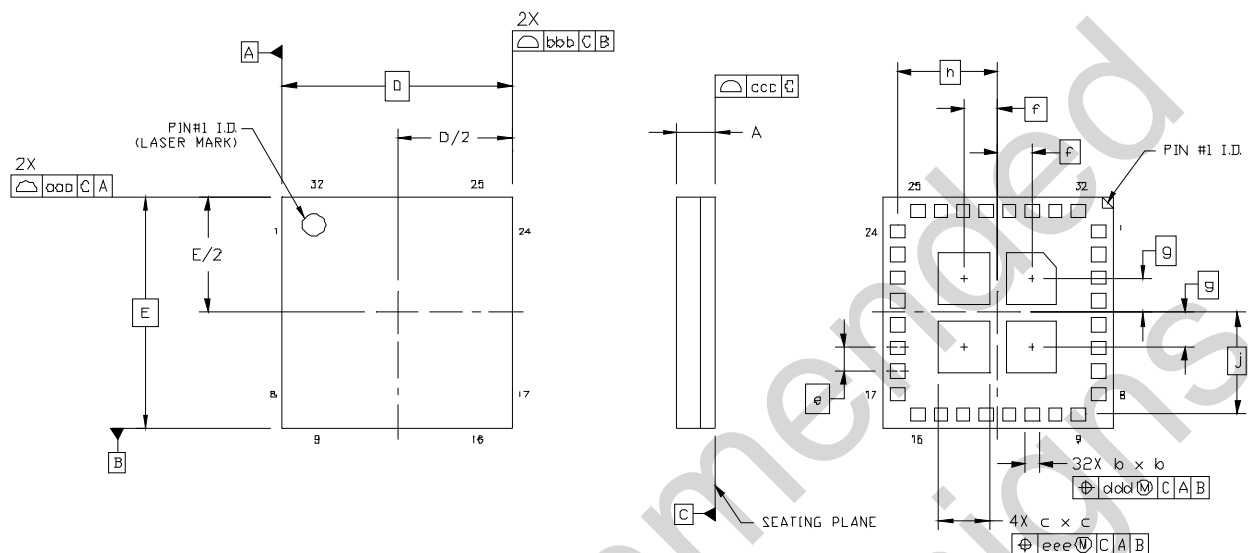
Figure 25. 32-Pin Land Grid Array Package (LGA)

Table 14. Package Diagram Dimensions

Dimension	Min	Nom	Max
A	0.75	0.85	0.95
b	0.27	0.30	0.33
c	1.00	1.10	1.20
D		5.00 BSC.	
e		0.50 BSC.	
E		5.00 BSC.	
f		0.735 BSC.	
g		0.735 BSC.	
h		2.185 BSC.	
j		2.185 BSC.	
aaa		0.10	
bbb		0.10	
ccc		0.10	
ddd		0.10	
eee		0.10	

Notes:
1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
3. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

17. Recommended VDD Power Supply Filtering

Because of the internal bypass capacitance and voltage regulators, the external supply bypass requirements for the Si5040 are minimal.

Note:
1. Place a .01 µF cap very close to each VDD pin (12, 15, 27, 32) and place a single 0.1 µF cap close to the Si5040.
2. No power sequencing is required.

Si5040
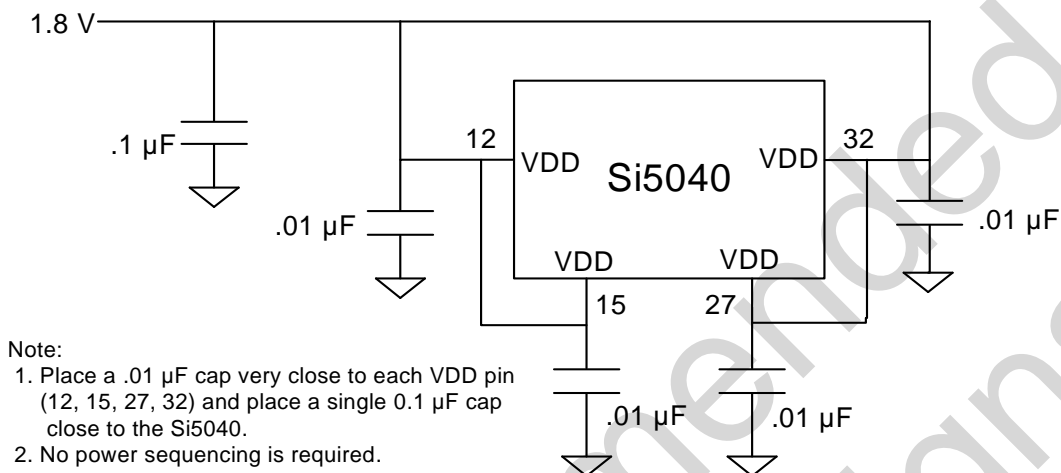
DOCUMENT CHANGE LIST

Revision 0.5 to Revision 0.8

- Updated final specification numbers for TBD items.
- Updated register name in Register 16.
- Changed aLosThresh[1:0] to aLosThresh[9:8] in Register 13.
- Jitter Tolerance measurement frequency changed from 400 MHz to 80 MHz.
- Corrected typos in the jitter transfer bandwidth specification in Tables 4 and 5.
- Corrected typos in Table 7, "CMU Timing Modes," on page 12.
- Updated crystal recommendation list.
- Removed I2C fall time spec in Table 8.
- Updated Constant Duty Cycle Control range to show adjustment range of 26% to 74% in Registers 21 and 22.
- Updated definition of Proportional Slice Threshold to show Threshold = 50% + sliceLvl/65536 x 100 in Registers 21and 22/
- Changed definition of Register 139, bit 4, from R/W to R.
- Clarified operation of fifoErr bit in the Register 4, bit 3.
- Updated definition of Reserved bits in Registers 56 and 184, bits 5:3.
- Reduced the input reference clock frequencies in Table 5 from 712.5 and 178.125 MHz to 709.38 and 177.34 MHz, respectively.
- Modified Detailed Block Diagram to show Slice_ADJ controlling the Programmable Equalizer block.
- Clarified definition of lolMode in Register 7 and 135, bit 2.
- Clarified description of Register 2, bit 4, as "SD pin drive configuration".
- Changed Reserved bit (Register 2, bit 7) to support configuration of RX_LOS pin as open drain output.
- The default configuration register bit settings changed to the following: Reg6[7:4] = 4h, Reg134[7:4] = 4h, Reg138[3:1] = 111b, Reg145[7:0] = 00h Reg154[7:2] = 1, Reg155[5:0] = 8, Reg184[7:6] = 11b, Reg56[7:6] = 11b, Reg85[7:5] = 111b.
- Revision ID, Register 1 bits 7:4, changed from 2 to 3 decimal to reflect Rev D.
- Updated Theta JA value in Table 11.
- Deleted Note 4 from Table 7.
- Updated "Acquisition Time" test condition (Register 68, bit3:0) in Table 3
- Removed both Stressed Eye Jitter Tolerance and Sinusoidal Jitter Tolerance from Table 3.

Revision 0.8 to Revision 0.85

- Corrected "OCIP2" typo in TXDOUT jitter test conditions in Table 5.
- Clarified jitter transfer peaking test condition in Table 5.
- Removed Note 2 from Table 5.
- Changed definition of bits 7:6 in Register 56 to Reserved.
- Changed definition of bits 7:6 in Register 184 to Reserved.
- Updated mechanical ground pad dimension in Table 14.
- Updated package drawing in Figure 25.
- Updated Table 14, "Package Diagram Dimensions," on page 104.

Revision 0.85 to Revision 0.86

- Updated title on page 1.
- Updated block diagram.
- Changed pins 10 and 11 to NC.
- Updated Typical Application Schematic (Section 3).
- Clarified operation of each slice mode in section "5.5 Receiver Slice Control".
- Updated Section "5.8. Receiver Loss of Lock (LOL)." Added technical info on RX SQM and Frequency LOL; added info on dynamic register control and acquisition time enhancement.
- Added Section "5.11. Recommended Pre-Emphasis on the RD Signal."
- Updated Section 6.4. Transmitter Loss of Lock (LOL). Added technical info on TX SQM and Frequency LOL; added info on dynamic register control and acquisition time enhancement.
- Removed section "6.7.4. Low Bandwidth Jitter Attenuation Mode (Mode 3)" and support for mode 3.
- Updated section "7. Loopback Modes" with technical info on XFI and Lineside loopback modes.
- Updated Section "12. Programmable Power Down Options."
- Added register summary and definitions in Section 13 for Registers 77, 98, 106-107, 205, and 226.
- Removed Section "18. Recommended Crystal Resonators" on page 94.
- Max power/current now specified for Mode 0 instead of Mode 3.

Revision 0.86 to Revision 1.2

- Removed sections 5.8.3 and 6.4.3 since fast acquisition is no longer supported.
- Clarified operation of Register 7[3:2] in register descriptions and "5.8.1. SQM LOL" on page 24.
- Updated part number in "15. Ordering Guide" on page 103.
- Added information on voltage application to VDD and VDDIO in pin descriptions.
- Removed reference to Mode 7 from Register 134 since Mode 7 does not exist.
- Changed all SCL references to SCK and all SDA reference to SD for consistency.

Revision 1.2 to Revision 1.3

- Clarified RX LOL feature in sections "5.8.1.1. Dynamic Register Control" and "5.8.2. Frequency LOL"
- Added section "5.8.3. Acquisition Time Enhancement"
- Clarified TX LOL feature in sections "6.4.1.1. Dynamic Register Control" and "6.4.2. Frequency LOL"
- Added section "6.4.3. Acquisition Time Enhancement"
- Changed TX and RX path data rate minimum parameters from 9.9 to 9.8 Gbps to allow for 9.8304 Gbps CPRI data rate

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